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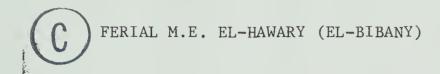




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THE UNIVERSITY OF ALBERTA TRANSIENT ANALYSIS OF TRANSMISSION LINES

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A THESIS

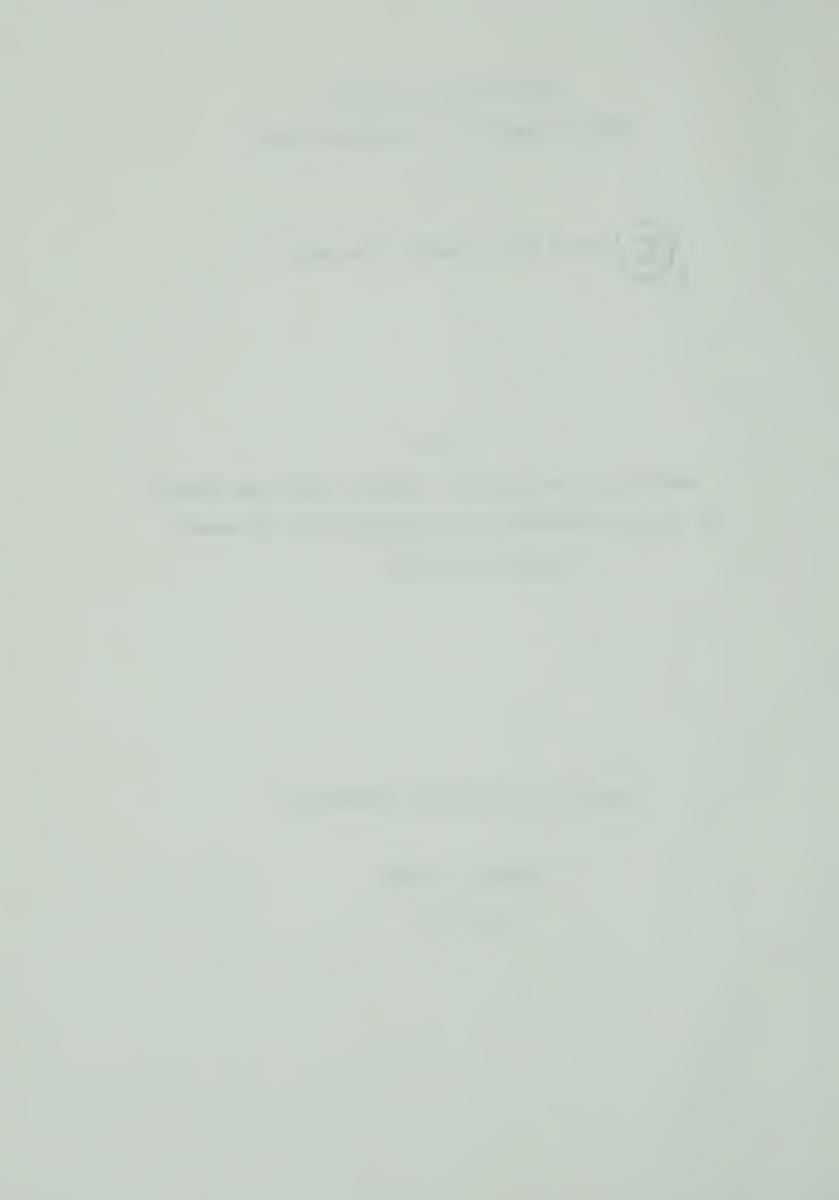
SUBMITTED TO THE FACULTY OF GRADUATE STUDIES AND RESEARCH
IN PARTIAL FULFILMENT OF THE REQUIREMENTS FOR THE DEGREE

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FACULTY OF GRADUATE STUDIES AND RESEARCH

The undersigned certify that they have read, and recommend to the Faculty of Graduate Studies and Research for acceptance, a thesis entitled "Transient Analysis of Transmission Lines" submitted by Ferial M.E. El-Hawary (El-Bibany) in partial fulfilment of the requirements for the degree of Master of Science.

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ABSTRACT

In this thesis, a method for the evaluation of the transient response of a nonuniform lossy transmission line is presented. The method of characteristics is employed to obtain difference equations describing the transmission line.

A stepped line approximation is used to analyze the transient response of the given nonuniform line. The concept of electrical length is employed in dividing the line into a number of equal delay sections. The set of difference equations describing the set stepped line is suitable for digital computer solution.

The computational procedure involving the use of a digital computer is illustrated for a specific distributions of L(x), C(x), r(x) and g(x).



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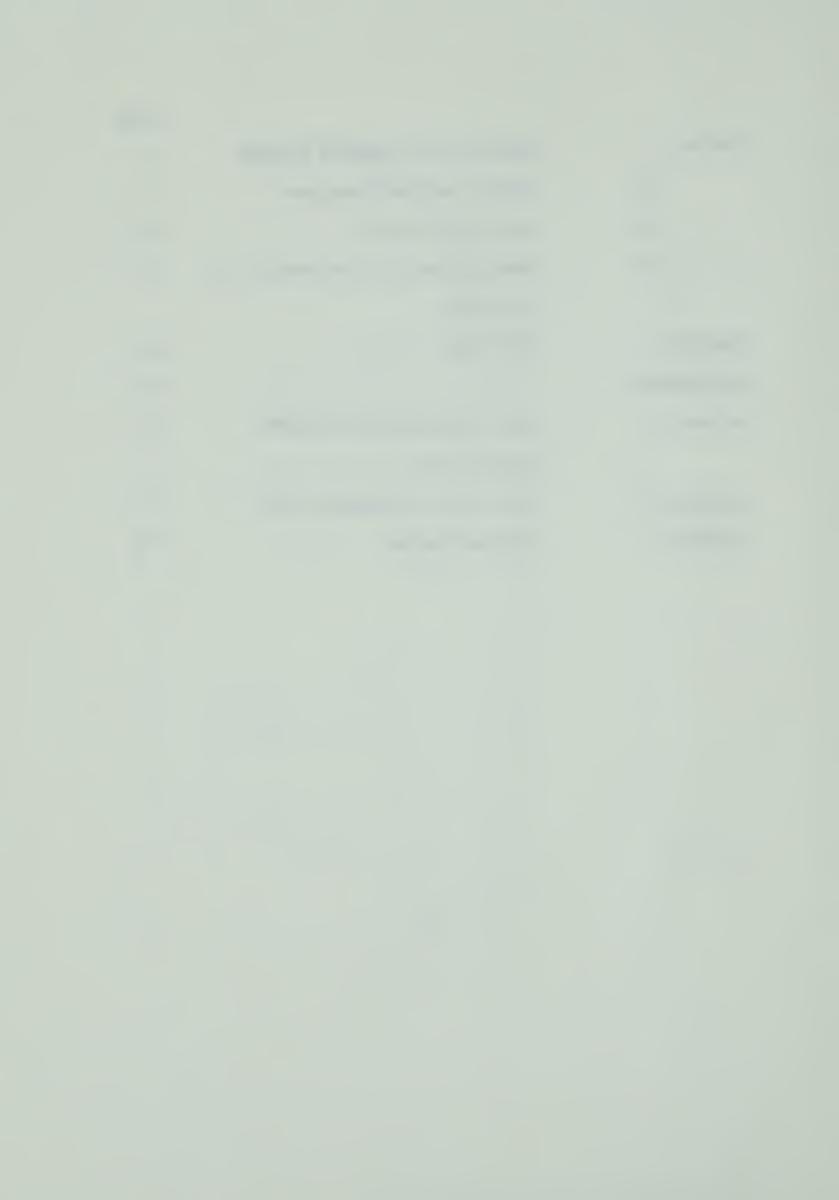


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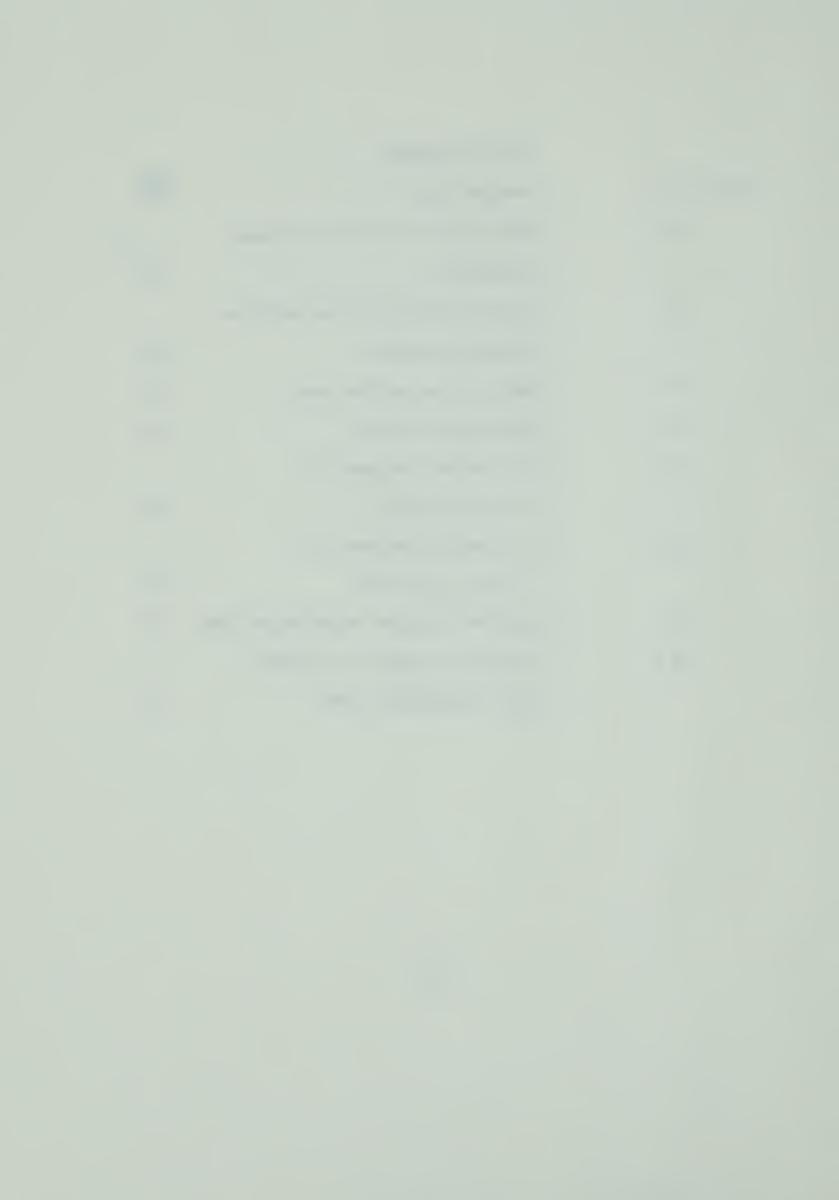
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CHAPTER-I

Introduction

1.1 Background

In the design of all apparatus and machines related to the generation and transmission of electrical power, it is important to know the salient features of the overvoltages likely to be encountered. Faults caused by switching are becoming more important with the ever higher transmission voltages being used.

Further the basic form of implementing computer hardware is in mounting integrated circuits (I.C.) packages on
multilayer printed circuit boards which in turn are mounted on
so called mother boards. The pins of the I.C. packages on the
printed circuit board are usually connected to each other by etched
metal strips, and circuit engineers are forced to regard these
metal strips as a length of a transmission line as the speed of
the circuit approaches the nanoseconds range (8).

Furthermore, the problem of constructing a mathematical model for simulating the electrical properties of memory arrays has been considered in 1963 by Weeks, (11) who found that a memory array can be approximated by a system of linear transmission lines.

Because of the above considerations, transient analysis of general transmission lines has recently become the object of



wide-spread interest of many investigators. The analysis of transient waves on a transmission line leads to a set of partial differential equations which define the relation between current and voltage as functions of time and position on the line. In general, these equations do not lend themselves to a closed form solution except in the special case of distortionless lines (where $\frac{R}{L} = G/C$), which includes the lossless case (R = G = 0). Thus using numerical techniques for solving such equations is inevitable

The application of the method of characteristics was shown in 1967 by F.H. Branin Jr., (1) to provide a simple analytic solution for the problem of a uniform lossless line. The method was shown to be superior in both speed and accuracy to the more familiar method of integrating the differential equations that describe a lumped L.C. model of the line. Branin did not attempt in this work to extend the method to more general lines.

Y.K. Liu, ⁽²⁾ in 1968, considered the transient analysis of a uniform lossy line terminated in a resistive load. For his analysis he used the method of characteristics and a second order Runge-Kutta technique to solve numerically the resulting pair of ordinary differential equations.

Transform methods for analyzing the transient response of uniform lines was considered by K.A. Chen in the same year. (3)

Chen's work was concerned with interconnected lines as an aid in the design of memory arrays in digital computer hardware.



The investigation carried out by Wassel (4) used transform methods for the analysis which was concerned with the effects of RC loadings of pulse signal transmission lines. It is noted that R. Murray-Shelley (5) investigated the same problem in the same year using the method of characteristics (often called the graphical method). In 1969, Y.K. Liu, (6) reported the application of the same technique he used in (2) when the line is terminated in a tunnel diode.

In 1970, V. Dvorak (9) reported a novel method of treating this problem. He used the method of characteristics in a way that is straight forward and avoids entirely lengthy numerical techniques such as Runge-Kutta method. One of the advantages of Dvorak's method is that any loading configuration can be handled easily. Further the treatment of nonuniform lines was shown to be best analyzed using a stepped line approximating the original line, that is the division of the line into a number of sections having the same delay (electrical length).

1.2 Scope of the thesis

In this thesis the application of the method of characteristics to the problem of computing the transient response of a general transmission line is presented. The analysis for a nonuniform loss-less line is presented. The method of analyzing both uniform and nonuniform lossy lines is discussed, here a stepped line approximating the given line is used. The choice of proper time step and



number of sections of the stepped line is considered for the case of uniform lines. This is to some extent an extension of the work of V. Dvorak. (9)

A general transmission line transient analysis computer program is described. The program is written in FORTRAN IV programming language. The program is discussed in some detail and numerical results are given for some example transmission lines.



CHAPTER 2

Transient analysis of lossless nonuniform transmission lines

2.1. The method of characteristics

The set of partial differential equations describing a uniform transmission line is,

$$-\frac{\partial V}{\partial Z} = L \frac{\partial i}{\partial t} + R i \qquad (2.1)$$

$$-\frac{\partial i}{\partial Z} = C \frac{\partial V}{\partial t} + G V \qquad (2.2)$$

where L, C, R and G are inductance, capacitance, resistance and conductance per unit length of the line respectively, V = V(Z,t) and i = i(Z,t) are the voltage and current, respectively, at distance Z from one end of the transmission line at time t. In essence the method of characteristics transforms (2.1) and (2.2) into two ordinary differential equations as follows. Let

$$\phi_1 = Ri + Li_t + V_Z = 0$$
 (2.3)

$$\phi_2 = GV + CV_t + i_Z = 0$$
 (2.4)



where

$$V_{Z} = \frac{\partial V}{\partial Z}$$

$$V_{t} = \frac{\partial V}{\partial t}$$

$$i_{Z} = \frac{\partial i}{\partial Z}$$

 $i_t = \frac{\partial i}{\partial t}$

Defining

$$\phi = \phi_1 + \lambda \phi_2 = 0$$

Then

$$\phi = Ri + \lambda GV + L[i_t + \frac{\lambda}{L}i_z] + \lambda C[V_t + \frac{V_z}{\lambda C}] = 0$$
 (2.5)

This is a voltage equation so that the quantity between brackets in the third term of (2.5) should be equal to the total derivative of the current $\frac{d\,\mathbf{i}}{d\,\mathbf{t}}$, hence we have

$$\frac{di}{dt} = i_t + \frac{\lambda}{L} i_Z$$

but

$$\frac{di}{dt} = i_t + \frac{\partial i}{\partial Z} \frac{dZ}{dt}$$

hence

$$\frac{\mathrm{dZ}}{\mathrm{dt}} = \frac{\lambda}{\mathrm{L}} \tag{2.6}$$



Similarly the quantity between brackets in the last term of (2.5) should be equal to the total derivative of the voltage $\frac{dV}{dt}$, hence

$$\frac{dZ}{dt} = \frac{1}{\lambda C} \tag{2.7}$$

by (2.6) and (2.7) we have

$$\lambda = \pm \sqrt{\frac{L}{C}}$$
 (2.8)

or

$$\frac{dZ}{dt} = \pm \frac{1}{\sqrt{LC}} \tag{2.9}$$

Substituting (2.8) into (2.5)

$$\phi = Ri \pm GV \sqrt{\frac{L}{C}} + L \frac{di}{dt} \pm \sqrt{LC} \frac{dV}{dt} = 0$$

or

$$\frac{d}{dt} \left[V \pm \sqrt{\frac{L}{C}} \quad i \right] = - \left[\frac{G}{C} V \pm \frac{R}{\sqrt{LC}} \quad i \right] \tag{2.10}$$

Now (2.9) implies that in the (Z-t) plane we have essentially straight line characteristic curves in the case of a uniform transmission line.

2.2 The stepped line approximation

Consider a lossless transmission line characterized by distributed inductance L(Z) > 0 and capacitance C(Z) > 0 per unit length, where Z is the physical position on the line. The electrical



position along the line is defined by:

$$y(Z) = \int_{0}^{Z} \sqrt{L(\eta) C(\eta)} d\eta$$
 (2.11)

and the local characteristic impedance by

$$\rho(y) = \sqrt{L[Z(y)]/C[Z(y)]}$$
 (2.12)

Now the characteristic curves in the (Z-t) plane are not straight lines in the case of nonuniform transmission lines, but the introduction of the electrical position y(Z) yields straight line characteristic curves in the (y-t) plane.

A transmission line of total length h may be divided into n sections of the same delay $\Delta y = \frac{y(h)}{n}$. If n is chosen large enough, then any such section can be approximated by a uniform lossless line with an average characteristic impedance ρ_K (1 < K < n) and with total propagation delay T = Δy . The resulting cascade of n uniform lossless terminating networks is described by 2 n difference equations in the time domain of the form given in (A-9) and (A-10), plus the two equations describing the terminating networks. A stepped line approximating the given nonuniform line is shown in figure 2.1.

The basic equations for a uniform lossless line are derived in Appendix A. Applying (A-9) to the incremental line between the $\left(K-1\right)^{\text{St}}$ and K^{th} nodes in figure 2.1 yields

$$V_{K}(t) + \rho_{K-1} i_{K}(t) = V_{K-1} (t-T) + \rho_{K-1} i_{K-1} (t-T)$$
 (2.13)

 $2 \le K \le n+1$



where we substituted

$$V_K(t) = V(1,t)$$

$$V_{K-1}(t) = V(0,t)$$

Further (A-10) is applied to the incremental line between the $K^{\mbox{th}}$ and ${(K-1)}^{\mbox{st}}$ nodes giving

$$V_{K}(t) - \rho_{K} i_{K}(t) = V_{K+1} (t-T) - \rho_{K} i_{K+1} (t-T)$$
 (2.14)
$$1 \le K \le n$$

The terminating networks A and B are ${f g}$ iven by their V-A characteristics

$$A : i_1 = f_1 (V_1)$$
 (2.15)

$$B : i_{n+1} = f_2(V_{n+1})$$
 (2.16)

Given the initial conditions at the time <code>instant t=0</code>, the values of V_K (t) and i_K (t) can then be computed at time instants t = mT, $1 \le m \le m_{max}$.



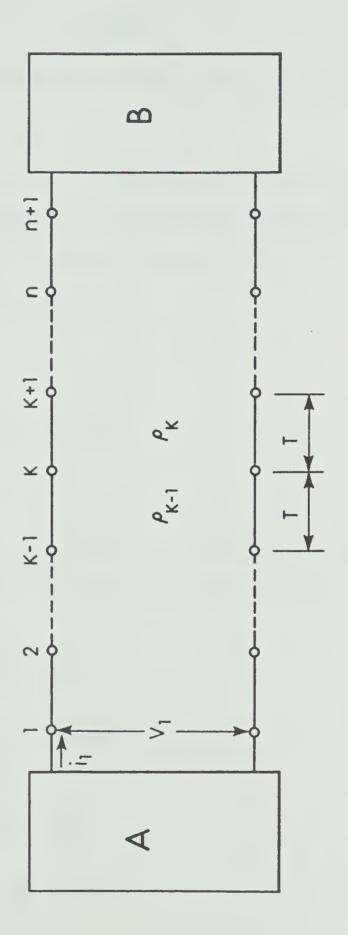


FIG. 2.1 A STEPPED LINE



CHAPTER 3

Transient analysis of lossy uniform lines

3.1 Analysis

Consider a lossy uniform line for which (2.9) implies that in the (Z-t) plane we have essentially straight line characteristic curves as shown in figure (3.1).

where along the α curve we have

$$\frac{dZ}{dt} = + \sqrt{\frac{1}{LC}}$$

while along the β curve we have

$$\frac{dZ}{dt} = -\frac{1}{\sqrt{LC}}$$

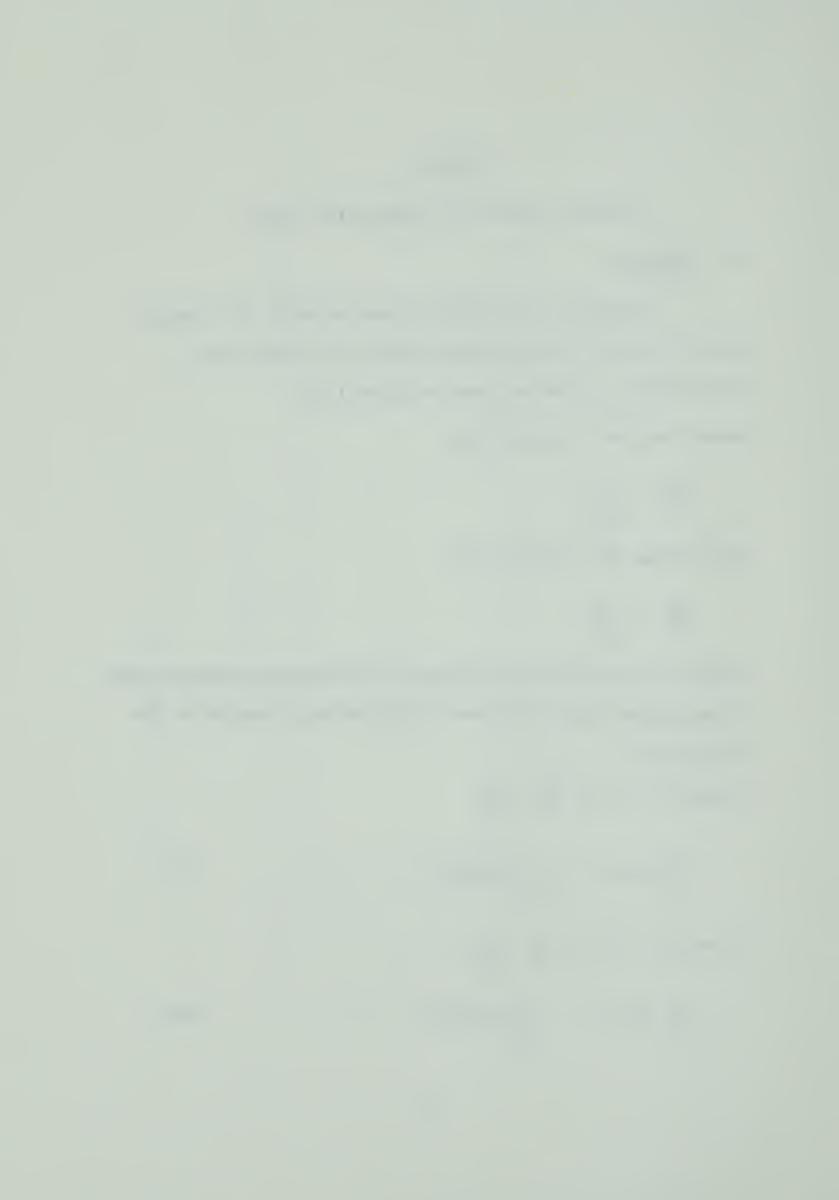
Using (2.10) we have the following two differential equations each being defined along a different characteristic direction in the (Z-t) plane

Along the α curve: $\frac{dZ}{dt} = \frac{1}{\sqrt{LC}}$

$$\frac{\mathrm{d}}{\mathrm{dt}} \left[V + \rho \mathbf{i} \right] = -\sqrt{\frac{1}{LC}} \left[\rho G V + R \mathbf{i} \right] \tag{3.1}$$

Along the β curve : $\frac{dZ}{dt} = \frac{-1}{\sqrt{LC}}$

$$\frac{\mathrm{d}}{\mathrm{d}t} \left[V - \rho i \right] = - \frac{1}{\sqrt{LC}} \left[\rho G V - R i \right]$$
 (3.2)



Note that (3.1) defines a forward propagating wave where as (3.2) defines a backward propagating wave.

Dividing the line into n sections of delay

$$T = h \sqrt{LC} / n \tag{3.3}$$

and taking $dt = \Delta t = T$, one gets

$$\Delta[V (Z,t) + \rho i (Z,t)] = \frac{-h}{n} [G \rho V (Z,t)]$$

K+1

+ R i (Z,t)]
$$\Delta[V (Z,t) - \rho i (Z,t)] = \frac{-h}{n} [G \rho V (Z,t)]$$
(3.4)

$$-Ri(Z,t)$$
] (3.5)

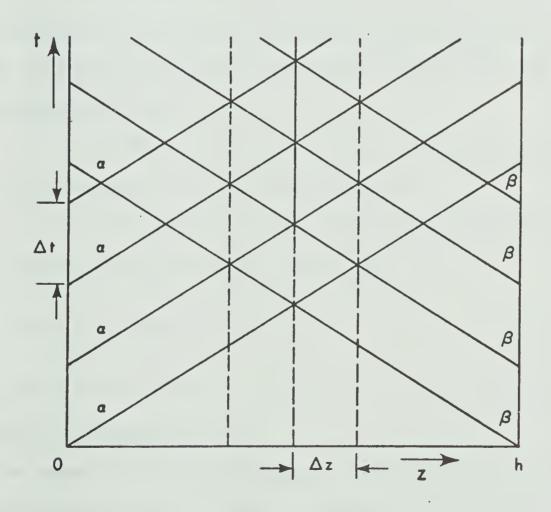


FIG. 3.1 FAMILIES OF α AND β CHARACTERISTIC CURVES IN Z-t PLANE



letting

$$g = \frac{Gh}{n}, \quad r = \frac{Rh}{n} \tag{3.6}$$

Then (3.4) and (3.5) become

$$\Delta[V (Z,t) + \rho i (Z,t)] = - [g \rho V (Z,t)]$$
+ r i (Z,t)] (3.7)

$$\Delta[V (Z,t) - \rho i (Z,t)] = - [g \rho V (Z,t)]$$

$$- r i (Z,t)]$$
(3.8)

Notice that any event occurring at the Kth node at the time instant (t-T) arrives at the (K+1)st node at the time instant, t, for the forward propagating wave, (the reverse is true for backward propagating wave).

So that writing (3.7) and (3.8) in a discrete (difference equation) form leads to the following question:

how are we going to discretize the right hand side of each of (3.7) and (3.8)? To this end we let

$$f(Z,t) = g \rho V(Z,t) + r i (Z,t)$$
 (3.9)

$$g(Z,t) = g \rho V (Z,t) - r i(Z,t)$$
 (3.10)

We have the following two cases

I. If we assume that f(Z,t) and g(Z,t) remains constant over the time interval [t-T, t], this is equivalent to the assumption

$$f(Z,t) = f_K(t-T)$$
 (3.11)



$$g(Z,t) = g_{K+1}(t-T)$$
 (3.12)

II. If we assume that f(Z,t) and g(Z,t) change in a linear fashion over the time interval [t-T,t] then

$$f(Z,t) = f_K(t-T) + \frac{f_{K+1}(t) - f_K(t-T)}{2}$$

or

$$f(Z,t) = [f_K(t-T) + f_{K+1}(t)] / 2$$
 (3.13)

and

$$g(Z,t) = [g_{K+1}(t-T) + g_K(t)] / 2$$
 (3.14)

Application of (3.11) and (3.12) to (3.7) and (3.8) respectively yields

$$V_{K+1}(t) + \rho i_{K+1}(t) = [1 - \rho g] V_{K}(t-T) +$$

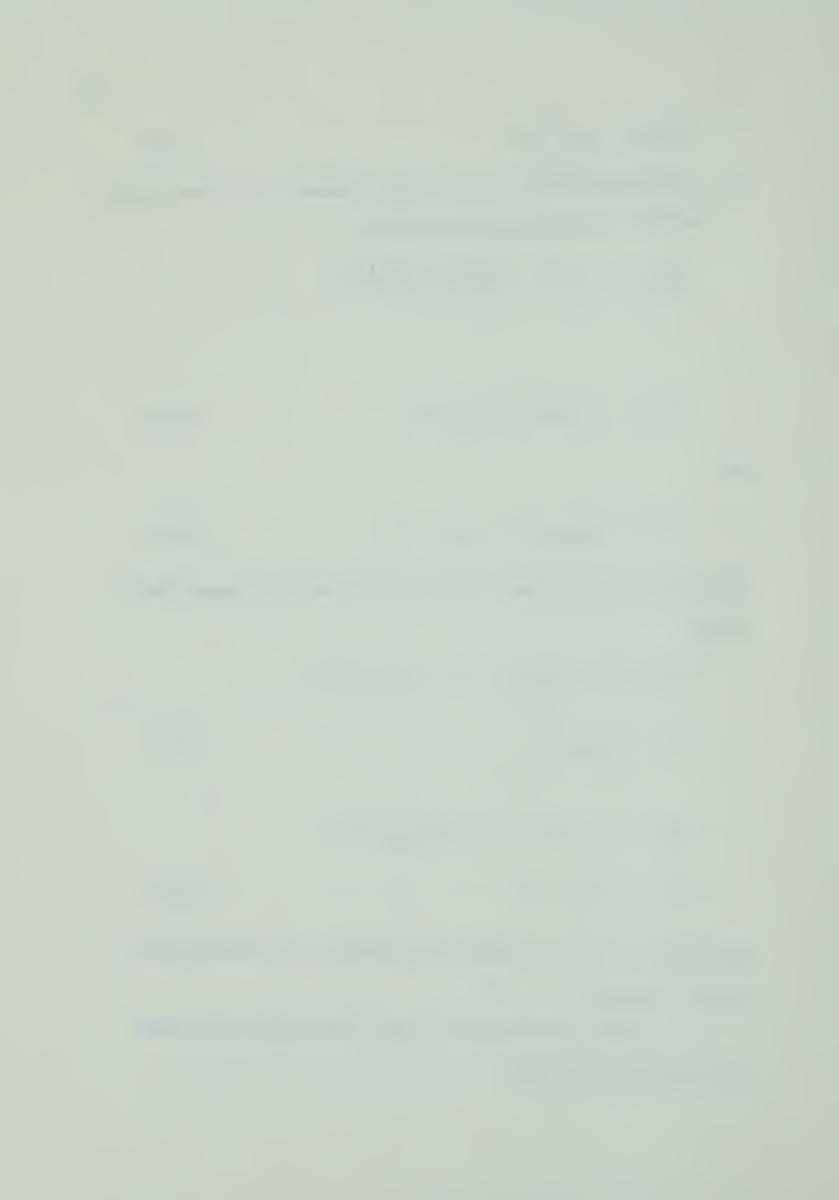
$$[\rho - r] i_{K}(t-T) \qquad (3.15)$$

$$V_{K}(t) - \rho i_{K}(t) = [1 - \rho g] V_{K+1}(t-T) -$$

$$[\rho - r] i_{K+1}(t-T)$$
(3.16)

Equations (3.15) and (3.16) will be referred to as the first formulation equations.

Further, application of (3.13) and (3.14) to (3.7) and (3.8) respectively yields



$$V_{K+1}(t) \left[1 + \frac{\rho g}{2}\right] + i_{K+1}(t) \left[\rho + \frac{r}{2}\right] =$$

$$V_{K}(t-T) \left[1 - \frac{\rho g}{2}\right] + i_{K}(t-T) \left[\rho - \frac{r}{2}\right]$$
(3.17)

$$V_{K}(t) \left[1 + \frac{\rho g}{2}\right] - i_{K}(t) \left[\rho + \frac{r}{2}\right] =$$

$$V_{K+1}(t-T) \left[1 - \frac{\rho g}{2}\right] - i_{K+1}(t-T) \left[\rho - \frac{r}{2}\right]$$
(3.18)

Equations (3.17) and (3.18) will be referred to as the second formulation equations.

The first and second formulation equations are the difference equations of a very short lossy line in the time domain.

3.2 Boundary conditions

Let the line have the following initial conditions

$$i_{K}(0) = 0$$
 $1 \le K \le n + 1$ (3.19)

$$V_{K}(0) = 0$$
 $1 \le K \le n + 1$ (3.20)

Let the sending end terminating network be an ideal current source whose current output is a prespecified function of time I(t). If the source is assumed to have a shunt conductance $G_{\rm S}$, then the following equation holds true.

$$i_1(t) = I(t) - G_S V_1(t)$$
 (3.21)

If the line is terminated in a load resistance $\boldsymbol{R}_{_{\boldsymbol{T}}}$ then

$$V_{n+1}(t) = R_{r-n+1}(t)$$
(3.22)

The line with its terminating network is shown in figure (3.2).



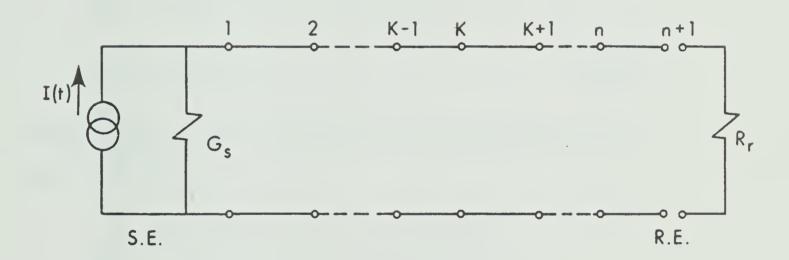


FIG. 3.2 UNIFORM TRANSMISSION LINE WITH

ITS TERMINATING NETWORKS



3.3 Implementing the first formulation

We rewrite the first formulation equations (3.15) and (3.16) as

$$V_{K}(t) + \rho i_{K}(t) = [1 - \rho g] V_{K-1}[t-T] +$$

$$[\rho - r] i_{K-1}(t-T) \qquad (3.23)$$

$$V_{K}(t) - \rho i_{K}(t) = [1 - \rho g] V_{K+1}(t-T) -$$

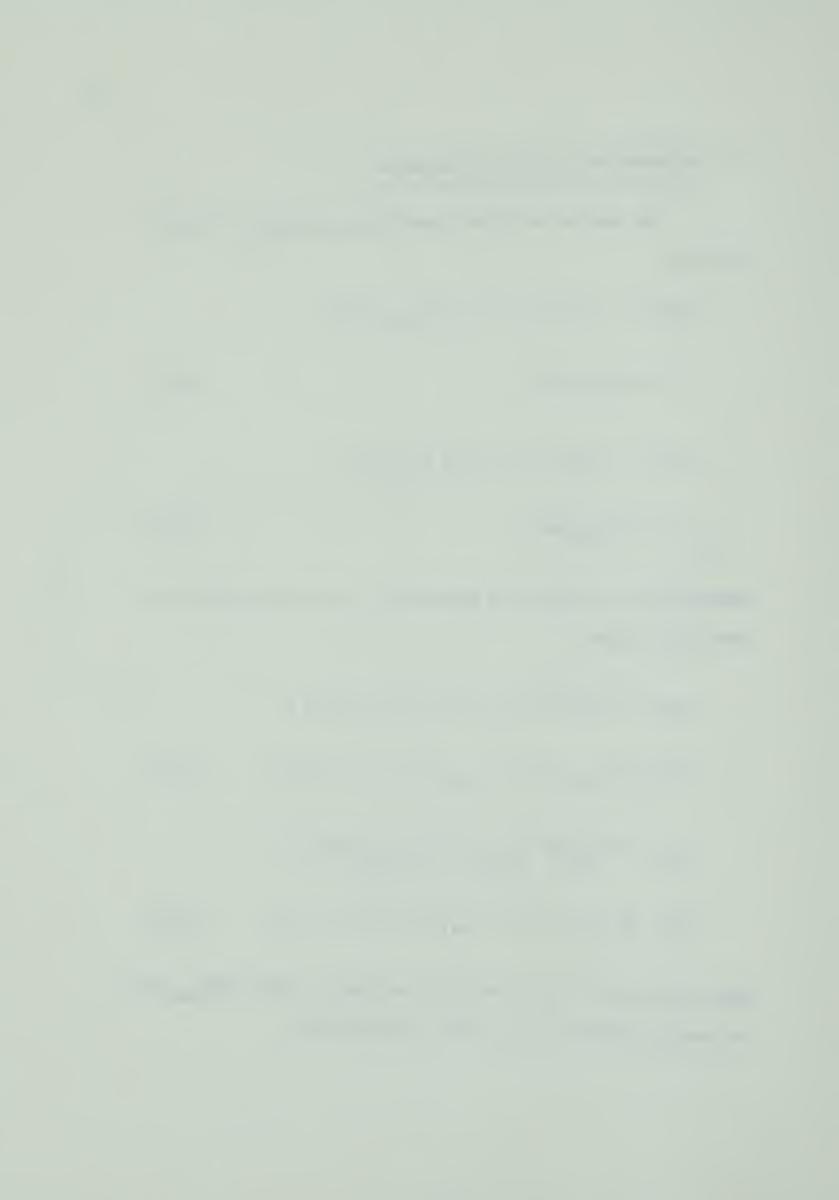
$$[\rho - r] i_{K+1}(t-T) \qquad (3.24)$$

Adding (3.24) to (3.23) and subtracting (3.24) from (3.23) respectively yields

$$V_{K}(t) = \frac{(1 - \rho g)}{2} \left[V_{K-1}(t-T) + V_{K+1}(t-T) \right] + \frac{(\rho - r)}{2} \left[i_{K-1}(t-T) - i_{K+1}(t-T) \right] + 2 \le K \le (n-1)$$

$$i_{K}(t) = \frac{(1 - \rho g)}{2\rho} \left[V_{K-1}(t-T) - V_{K+1}(t-T) \right] + \frac{i_{K}(t-T)}{2\rho} \left[i_{K-1}(t-T) + i_{K+1}(t-T) \right] + \frac{i_{K}(t-T)}{2\rho} \left[i_{K-1}(t-T) + i_{K-1}(t-T) \right] + \frac{i_{K}(t-T)}{2\rho} \left[i_{K-1}(t-T) + i_{K-1}(t-T) \right] + \frac{i_{K}(t-T)}{2$$

Thus the values of $V_K(t)$ and $i_K(t)$ at times t = mT, $1 \le m \le m_{max}$ can be computed using (3.23), (3.24), (3.25) $\sqrt[4]{(3.26)}$.



3.4 Implementing the second formulation

Rewriting the second formulation equations (3.17) and (3.18) as

$$V_{K}(t) \left[1 + \frac{\rho g}{2}\right] + i_{K}(t) \left[\rho + \frac{r}{2}\right] = V_{K-1}(t-T)\left[1 - \frac{\rho g}{2}\right]$$

$$+ i_{K-1}(t-T) \left[\rho - \frac{r}{2}\right] \qquad (3.27)$$

$$V_{K}(t) \left[1 + \frac{\rho g}{2}\right] - i_{K}(t) \left[\rho + \frac{r}{2}\right] = V_{K+1}(t-T)\left[1 - \frac{\rho g}{2}\right]$$

$$- i_{K+1}(t-T) \left[\rho - \frac{r}{2}\right] \qquad (3.28)$$

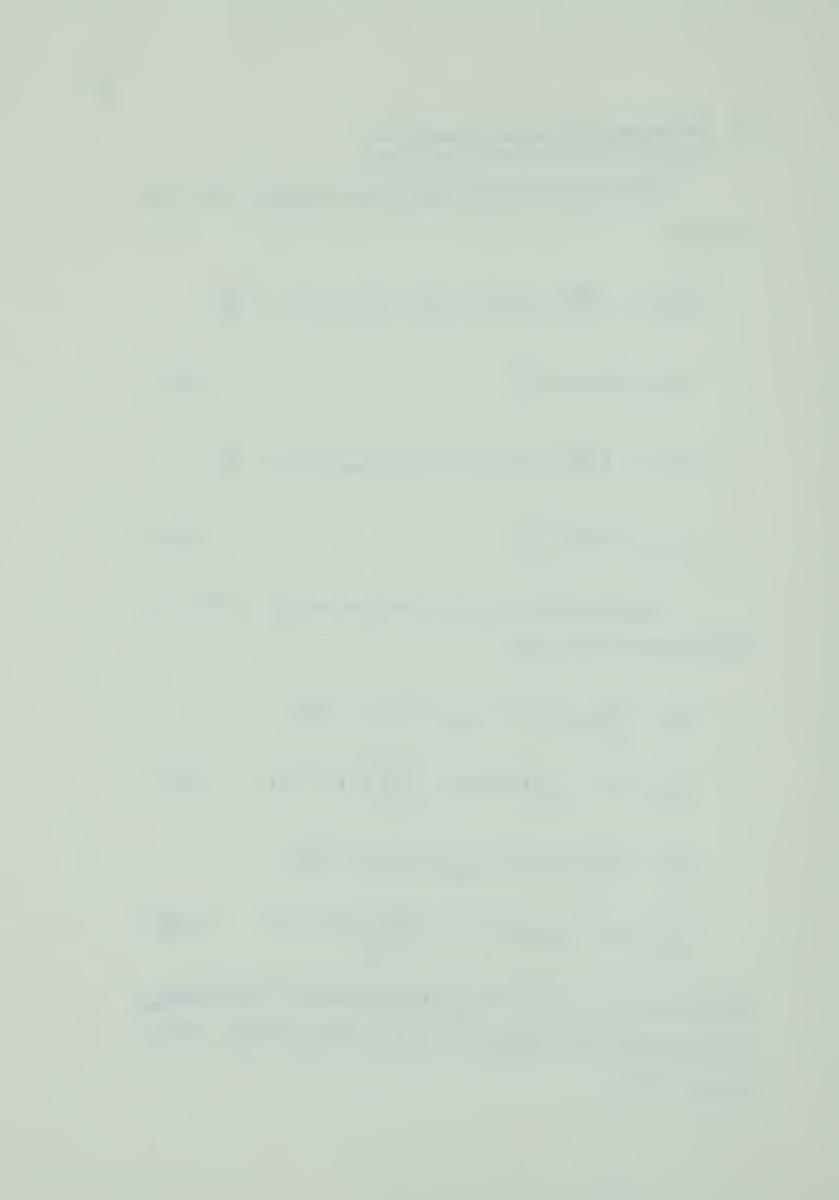
Adding (3.27) and (3.28) and subtracting (3.28) from (3.27) respectively yields

$$V_{K}(t) = \left\{ \left[V_{K-1}(t-T) + V_{K+1}(t-T) \right] \left[1 - \frac{\rho g}{2} \right] + \left[i_{K-1}(t-T) - i_{K+1}(t-T) \right] \left[\rho - \frac{r}{2} \right] \right\} / (2 + \rho g)$$

$$i_{K}(t) = \left\{ \left[V_{K-1}(t-T) - V_{K+1}(t-T) \right] \left[1 - \frac{\rho g}{2} \right] + \left[i_{K-1}(t-T) + i_{K+1}(t-T) \right] \left[\rho - \frac{r}{2} \right] \right\} / (2\rho + r)$$

$$(3.30)$$

Thus the values of $V_K(t)$ and $i_K(t)$ at instants t = mT, $1 \le m \le m_{max}$ can be computed using equations (3.19), (3.20), (3.29), (3.30), (3.21), (3.22).



3.5 Error estimation

If we let

$$W(Z,t) = V(Z,t) + \rho i(Z,t)$$
 (3.31)

$$f(Z,t) = \frac{-1}{\sqrt{LC}} [G \rho V(Z,t) + R i(Z,t)]$$
 (3.32)

Then equation (3.1) is transformed to

$$\frac{dW}{dt} = f(t) \tag{3.33}$$

for a specific position Z.

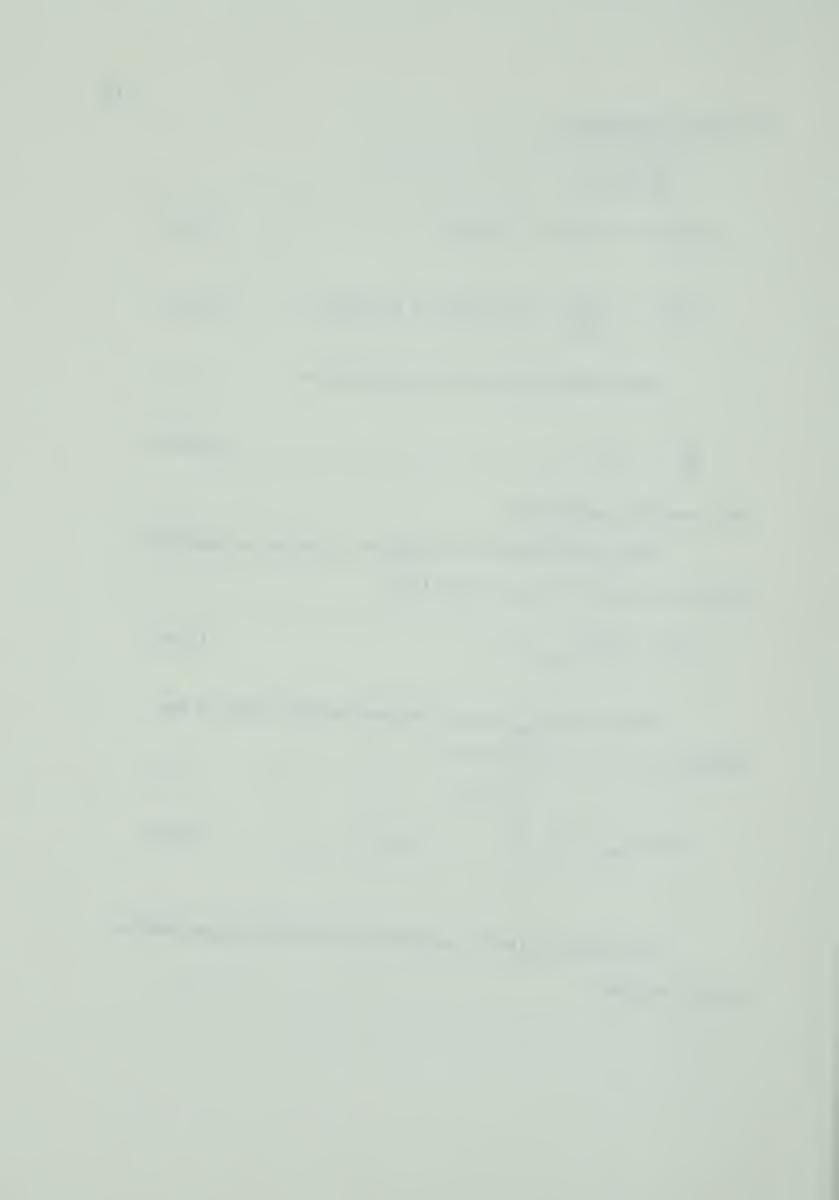
Thus the increment of change of W, which is denoted by ΔW over the interval [t_o, t_o + Δt] is

$$\Delta W = [f(t)]_{avg} \cdot \Delta t$$
 (3.34)

Where $[f(t)]_{avg}$ is the average value of f(t) in the interval $[t_o, t_o + \Delta t]$ given by

$$[f(t)]_{avg} = \frac{1}{\Delta t} \int_{0}^{t} f(s) ds$$
 (3.35)

Now using Taylor's expansion for the right hand side of (3.35) one gets



$$[f(t)]_{avg} = f(t_0) + \frac{\Delta t}{2!} \dot{f}(t) \Big|_{t_0} + \frac{(\Delta t)^n}{3!} \dot{f}(t) \Big|_{t_0} + \dots$$

$$(\frac{\Delta t}{3!})^2 \dot{f}(t) \Big|_{t_0} + \dots + \frac{(\Delta t)^n}{(n+1)!} \dot{f}^n \Big|_{t_0} + \dots$$
(3.36)

If Δt is sufficiently small then

$$[f(t)]_{avg} = f(t_0) + \frac{\Delta t}{2!} \dot{f}(t) \Big|_{t_0} + \frac{(\Delta t)^2}{3!} \dot{f}(t) \Big|_{t_0} + \frac{(\Delta t)^3}{4!} \dot{f}(t) \Big|_{t_0}$$
(3.37)

Thus a sufficiently accurate value of ΔW is

$$\Delta W = \Delta [f_0 + \frac{\Delta}{2} f_0 + \frac{\Delta^2}{6} f_0 + \frac{\Delta^3}{24} f_0]$$
 (3.38)

where

$$f_{o} = f(t_{o})$$

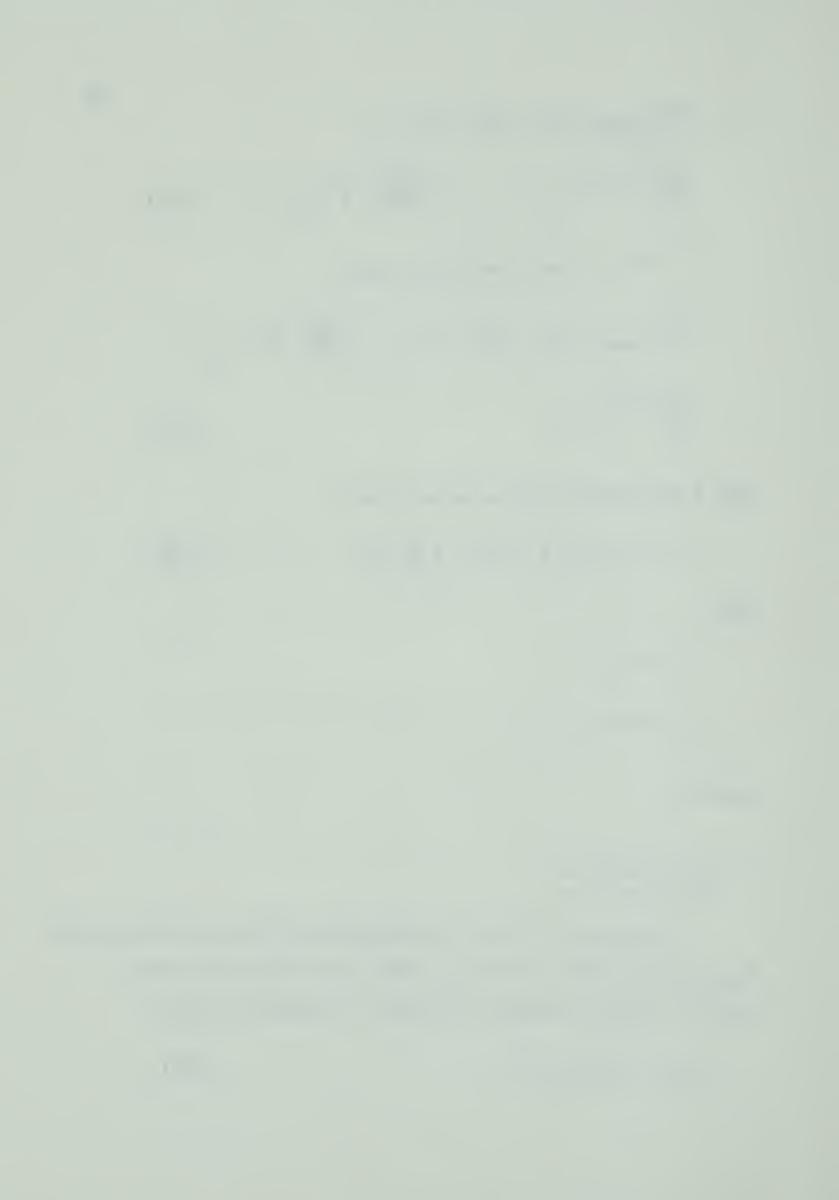
$$f_{o} = f(t) \mid_{t_{o}}$$

and so on.

a - First formulation

Referring to (3.11), the average value of f(t) over the interval $[t_o, t_o^{+\Delta}]$ was taken to be $f(t_o)$, (where $t_o^{=t-T}$, $\Delta=T$ in this case) thus the estimated increment of change in W, denoted by ΔW_{s1} is

$$\Delta W_{sl} = [f(t)]_{avg_1}. \Delta t$$
 (3.39)



hence

$$\Delta W_{S_1} = f_0 \cdot \Delta \tag{3.40}$$

This provides us with a sufficiently accurate error measure for the first formulation which we denote by $\epsilon_{\rm sl}$, so we have

$$\varepsilon_{s_1} = |\Delta W_{s_1} - \Delta W| \qquad (3.41)$$

Substituting (3.38) and (3.40) into (3.41) we get

$$\varepsilon_{s_1} = \frac{\Delta^2}{2} f_0 + O_3 (\Delta t)$$
 (3.42)

with $0_3(\Delta t)$ being terms in Δt higher than third order.

Now if an upper limit on the error incurred in implementing the first formulation is given by $\boldsymbol{\epsilon}_a$, then,

$$\varepsilon_a > \frac{\Delta^2}{2}$$
 for

or

$$\Delta t < \left(\frac{2\varepsilon_a}{f}\right)^{\frac{1}{2}} \tag{3.43}$$

which defines an upper limit on the value of the time step to be chosen so that the given accuracy is achieved. Let this upper limit be Δt_1 , then

$$\Delta t_1 = \begin{pmatrix} \frac{2\varepsilon_a}{f} \end{pmatrix}^{\frac{1}{2}}$$



b - Second formulation

Referring to (3.13), the average value of f(t) over the interval $[t_0, t_0^{+\Delta}]$ was taken to be

$$[f(t)]_{avg_2} = \frac{1}{2}[f_0 + f_1]$$
 (3.44)

where

$$f_1 = f(t_0 + \Delta t) \tag{3.45}$$

thus the estimated increment of change in W denoted by ΔW_{s_2} is

$$\Delta W_{s_2} = [f(t)]_{avg_2}. \quad \Delta t$$
 (3.46)

which yields the error measure for the 2nd formulation $\epsilon_{\mbox{\scriptsize s}_2}$ which is given by

$$\varepsilon_{s_2} = |\Delta W_{s_2} - \Delta W| \qquad (3.47)$$

applying (3.38) and (3.46) to (3.47) yields

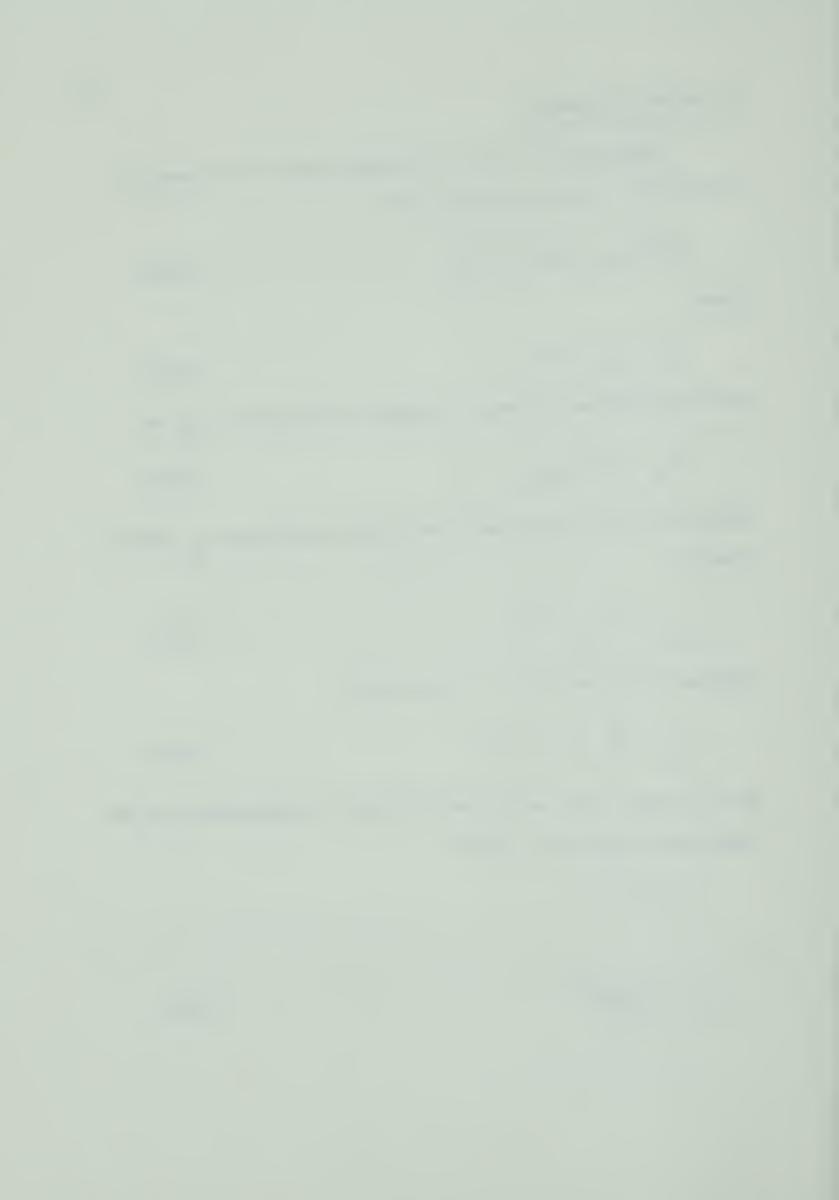
$$\varepsilon_{s_2} = \frac{\Delta^3}{12} + 0_4 (\Delta t)$$
 (3.48)

Now if an upper limit on the error incurred in implementing the 2nd formulation is given by $\boldsymbol{\epsilon}_a$ then

$$\epsilon_a > \frac{\Delta^3}{12} f_o$$

or

$$\Delta t < \left(\frac{12\varepsilon_a}{\tilde{f}_a}\right)^{1/3} \tag{3.49}$$



define

$$\Delta t_2 = \begin{pmatrix} \frac{12\varepsilon_a}{3} & \frac{1}{3} \\ \frac{1}{3} & \frac{1}{3} \end{pmatrix}$$

This specifies the upper limit (Δt_2) on the value of the time step to be chosen, so that the given accuracy is achieved.

In appendix (B) it is shown that the function W obeys the following equation

$$W_{K+1} = W_K \exp[-\Delta t \cdot s - \frac{1}{2} (\frac{r}{\rho} + g\rho)]$$
 (3.50)

where s is the Laplace operator and hence the term $\Delta t.s$ in the exponent represents the time delay between two consecutive nodes.

We are concerned only with the deterioration in the wave W, not the delay, so that we need only to consider

$$W_{K+1} = W_K \exp\left[-\frac{1}{2}\left(\frac{r}{\rho} + g\rho\right)\right]$$
 (3.51)

Let

$$\varepsilon = \exp\left[-\frac{1}{2}\left(\frac{\mathbf{r}}{\rho} + g\rho\right)\right] \tag{3.52}$$

Then

$$W_{K+1} = \varepsilon W_K \tag{3.53}$$

$$W_{K+2} = \varepsilon^2 W_K \tag{3.54}$$

$$W_{K+3} = \varepsilon^3 W_K \tag{3.55}$$



This enables one to evaluate f and f which are required in the calculation of upper limits on the time step in the previous section.

By definition

$$f(t) = \frac{dW}{dt} \tag{3.56}$$

Then in a discrete form one can write

$$f_{o} = \frac{W_{K+1} - W_{K}}{\Delta t} \tag{3.57}$$

Substituting (3.53) into (3.57) then

$$f_{o} = \frac{W_{K}}{\Delta t} \quad (\varepsilon - 1) \tag{3.58}$$

where

$$f_{o} = f(o)$$
 (3.59)

and

$$f_1 = \frac{W_{K+2} - W_{K+1}}{\Delta t}$$
 (3.60)

Substituting (3.53) and (3.54) into (3.60) one gets

$$f_1 = \frac{W_K}{\Delta t} \quad \varepsilon (\varepsilon - 1) \tag{3.61}$$

Similarly

$$f_2 = \frac{W_K}{\Lambda t} \epsilon^2 (\epsilon - 1) \tag{3.62}$$

where

$$f_{m} = f(m.\Delta t) \tag{3.63}$$



further

$$f(t) = \frac{d f(t)}{dt}$$
 (3.64)

in a discrete form

$$f_{o} = \frac{f_{1} - f_{o}}{\Delta t} \tag{3.65}$$

hence using (3.58) and (3.61)

$$f_{o} = \frac{W_{K}}{(\Delta t)^{2}} (\varepsilon - 1)^{2}$$
(3.66)

and

$$f_1 = \frac{W_K}{(\Delta t)^2} \quad \varepsilon (\varepsilon - 1)^2 \tag{3.67}$$

where

$$f_{m} = f(t) \mid_{(m,\Delta t)}$$
 (3.68)

since

$$f(t) = \frac{d f(t)}{dt}$$
 (3.69)

in a discrete form

$$f_{O} = \frac{\dot{f}_{1} - \dot{f}_{O}}{\Delta t} \tag{3.70}$$

hence, using (3.66) and (3.67) one gets

$$f_{o} = \frac{W_{K}}{(\Delta t)^{3}} (\varepsilon - 1)^{3}$$
(3.71)



d - proper number of sections

The results obtained in the last two sections will now be applied to obtain an estimate of the minimum number of sections to which the line should be divided in order to meet a specified accuracy limit.

Let the total allowable error be ΔW_{t} defined as:

$$\Delta W_{t} = n \cdot \epsilon_{a}$$
 (3.72)

Denote the relative error in calculating W by μ so that

$$\mu = \frac{\Delta W_{t}}{W_{K}} \tag{3.73}$$

Also we have by (3.52)

$$\varepsilon = \exp\left[-\frac{\alpha}{n}\right] \tag{3.74}$$

where

$$\alpha = \frac{1}{2} \left[\frac{R}{\rho} + G \rho \right] \tag{3.75}$$

In the practical case where $r<<\rho$ and $g<<\frac{1}{\rho},~\frac{\alpha}{n}<<1$ and ϵ can be approximated by

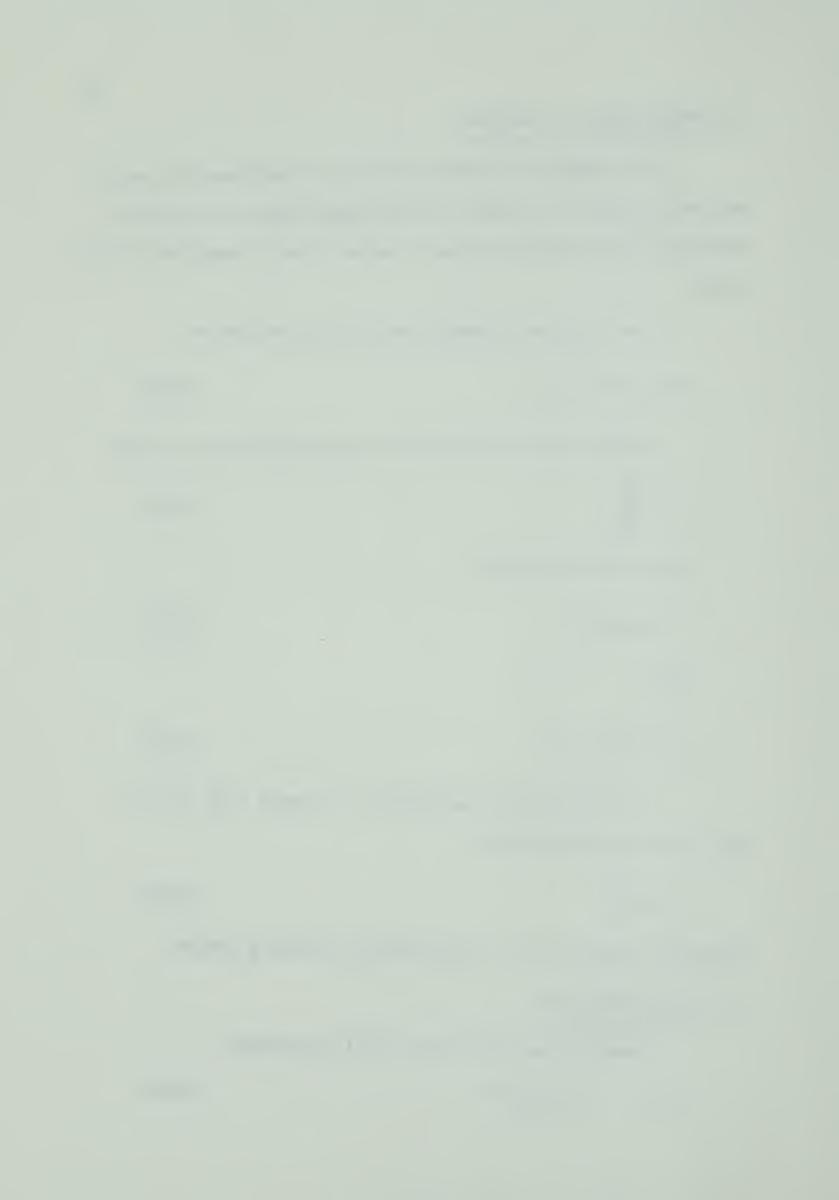
$$\varepsilon = 1 - \frac{\alpha}{n} \tag{3.76}$$

Using this expression for ϵ we obtain the following results.

A - First formulation

Substituting (3.66) into (3.43) one obtains

$$(1-\varepsilon) < (2\varepsilon_a/W_K)^{\frac{1}{2}}$$
 (3.77)



further substitute (3.72) and (3.73) into (3.77) to obtain

$$(1-\varepsilon) < \sqrt{\frac{2\mu}{n}}$$
 (3.78)

Using (3.76) in (3.78) one obtains

$$\frac{\alpha}{n} < \sqrt{\frac{2\mu}{n}} \tag{3.79}$$

or

$$n > \frac{\alpha^2}{2\mu} \tag{3.80}$$

Thus the minimum number of sections to be chosen for the first formulation is given by

$$n_1 = \frac{\alpha^2}{2\mu} \tag{3.81}$$



B - Second formulation

Substituting (3.71) into (3.49) one obtains

$$1-\varepsilon < \left(\frac{12\varepsilon_a}{W_K} \right)^{1/3} \tag{3.82}$$

We obtain upon substitution of (3.72) and (3.73) into (3.82) we get

$$1-\varepsilon < \left(\frac{12\mu}{n}\right)^{1/3} \tag{3.83}$$

Then (3.76) and (3.83) yield

$$\frac{\alpha}{n} < \left(\frac{12\mu}{n}\right) \tag{3.84}$$

or

$$n > \left[\alpha^3/12\mu\right]^{\frac{1}{2}}$$
 (3.85)

Thus the minimum number of sections to be chosen for the second formulation is given by

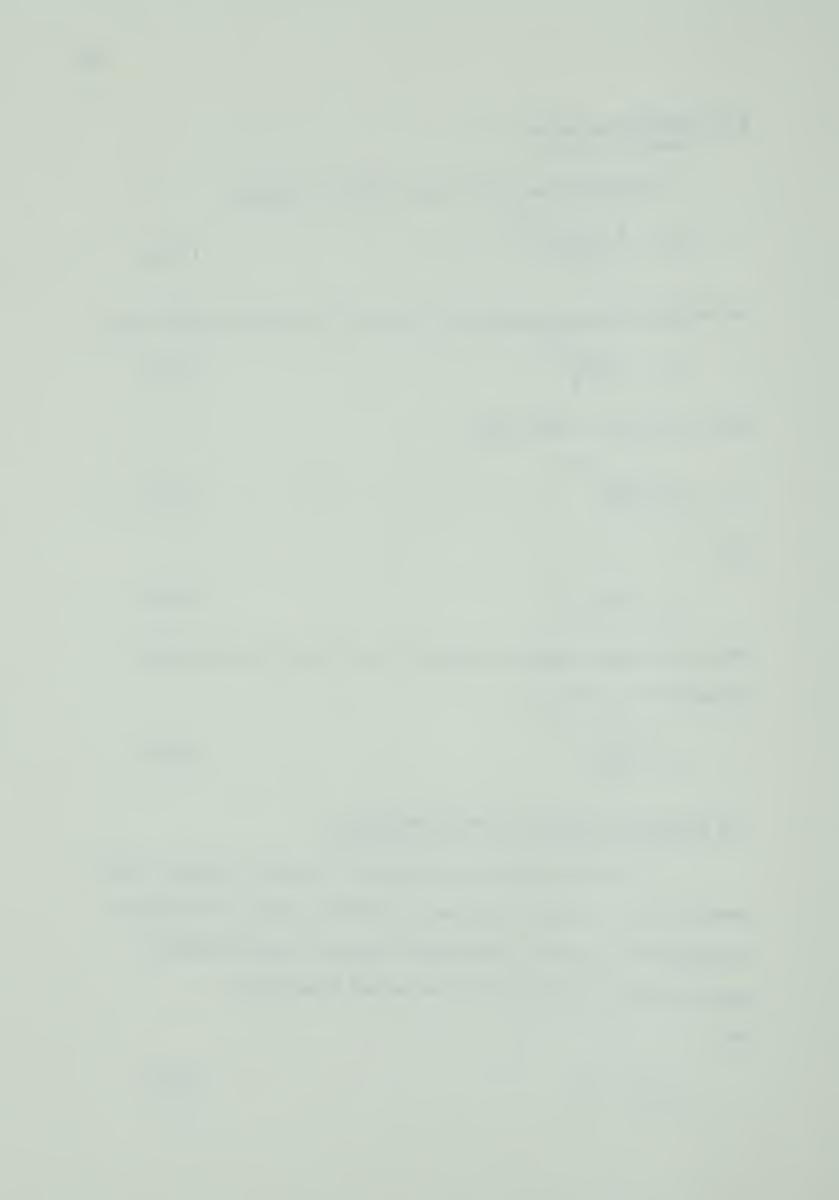
$$n_2 = \left(\frac{\alpha^3}{12\mu}\right)^{\frac{1}{2}} \tag{3.86}$$

3.6 Comparison between the two formulations

In this section it is shown that for most practical transmission lines to achieve the same accuracy by using the two given formulations, the first formulation requires a larger number of sections than that required by the second formulation.

Let

$$p = n_1^2 - n_2^2 (3.87)$$



where n_1 and n_2 are the minimum number of sections for a given accuracy measure μ required for the first and second formulations respectively as given by (3.81) and (3.86).

Hence

$$p = \frac{\alpha^3}{4\mu} \left[\frac{\alpha}{\mu} - \frac{1}{3} \right]$$
 (3.88)

The required relative error normally does not exceed 0.01, thus we can assume

$$\mu < 0.01$$
 (3.89)

Now by definition

$$\alpha = \frac{1}{2} \left[\frac{R}{\rho} + G\rho \right] \tag{3.90}$$

so that using (3.89) and (3.90)

If

$$\frac{R}{\rho} + G \rho > \frac{1}{150}$$
 (3.91)

then

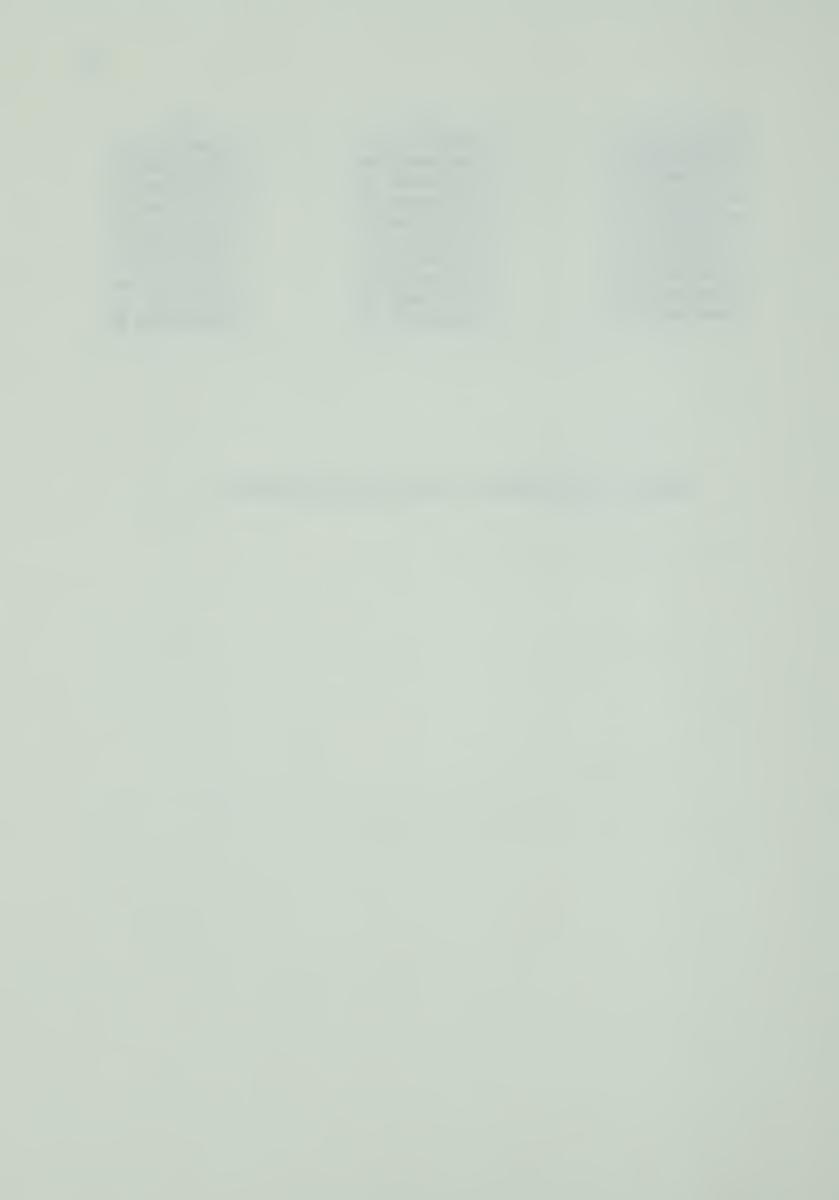
$$p > 0$$
 and $n_1 > n_2$

table 3.1 gives a comparison between \mathbf{n}_1 and \mathbf{n}_2 for different values of $\alpha.$



α	n ₁	n ₂
0.99999964E-01	0.49999967E-01	0.28867505E-01
0.19999993E 00	0.1999 99 87E 00	0.81649601E-01
0.29999989E 00	0.44999969E 00	0.14999998E 00
0.39999986E 00	0.79999948E 00	0.23094004E 00
0.49999987E 00	0.12499990E 01	0.32274854E 00
0 59999979E 00	0.17999983E 01	0.42426401E 00
0.69999975E 00	0.24499979E 01	0.53463376E 00
0-79999971E 00	0.319999 7 9E 01	0.65319 7 17E 00
0.89999968E 00	0-40499983E 01	0.77942276F 00
0-99999964E 00	0.49999981E 01	0.91287082E 00

TABLE 3.1 VARIATION OF n_1 AND n_2 WITH α FOR μ =0.1



CHAPTER 4

TRANSIENT ANALYSIS OF LOSSY NONUNIFORM LINES

4.1 Dividing the line

Consider a lossy transmission line characterized by distributed inductance L(Z) > 0, capacitance C(Z) > 0, resistance R(Z) > 0 and conductance G(Z) > 0 per unit length, where Z is the physical position on the line. In this case the relations given by equations (3.1) and (3.2) hold true, but with the transmission line parameters varying with Z. This means that the characteristic curves in the Z-t plane are no longer straight lines as is the case for uniform lines.

The electrical position along the line is defined by

$$Y(Z) = \int_{0}^{Z} \sqrt{L(\xi) C(\xi)} d\xi$$
 (4.1)

From this definition it is evident that the characteristic curves in the Y-t plane are straight lines. It is worth noting here that the electrical position is the time delay that a wave initiated at Z=O takes to arrive at the physical position Z.

Thus if we have a line whose length is h, the total time delay would be

$$\zeta = \int_{0}^{h} \sqrt{L (\xi) C (\xi)} d\xi \tag{4.2}$$

Now if the physical line length was divided into any number of equal sections, the electrical length (delay) of these sections will



not be equal. Hence we need to divide the line into n sections of equal electrical length, the method that will be used is as follows:

I — Divide the physical line length h into N sections, so that each section is of physical length $\frac{h}{N}$.

Thus we have created an N-component vector $[Z_K]$ representing the physical length between the sending end and the $(K+1)^{\text{st}}$ node. With $Z_K = (K-1) \cdot \frac{h}{N}$ $K=2,\ldots,$ N+1 (4.3)

II - Using (4.1), another N-component vector $[Y_K]$ representing the electrical length between the sending end and the (K+1)st node corresponding to the above divisions.

III - Now the electrical line length ζ is divided into n equal sections. The delay of each is Δt .

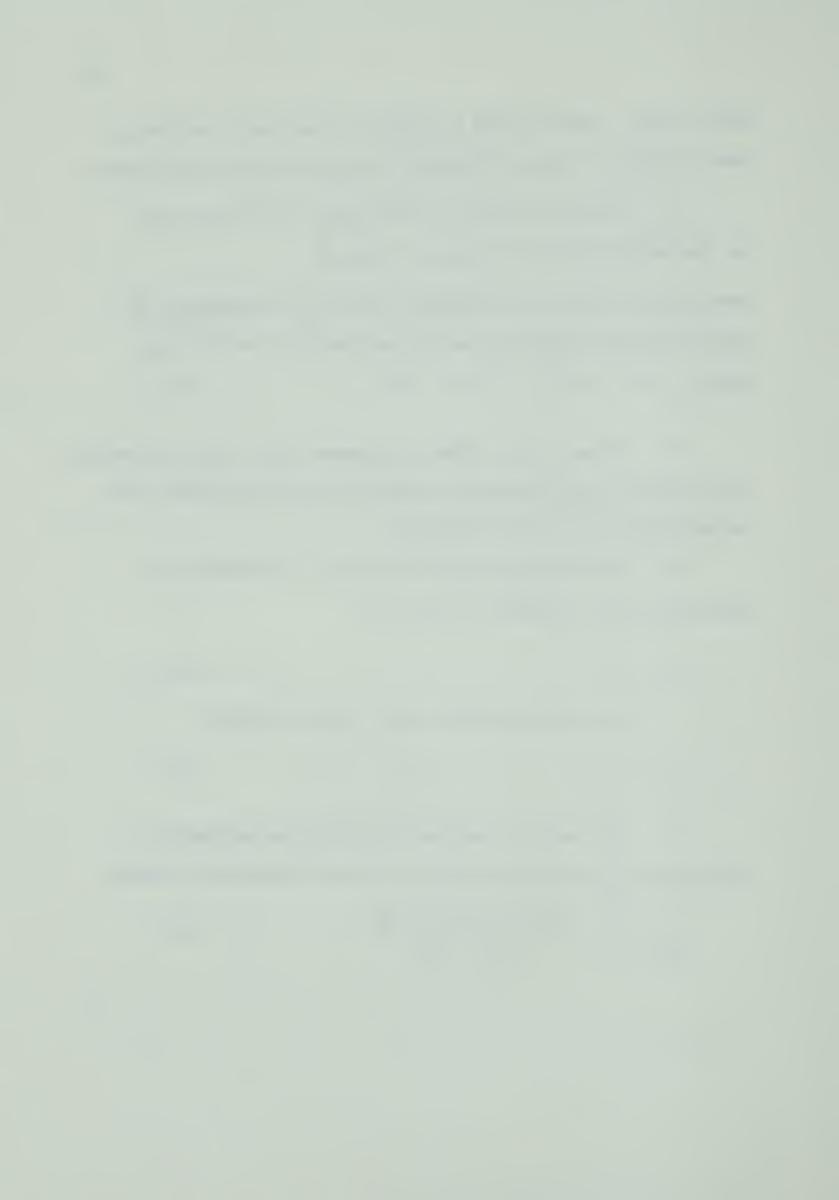
$$\Delta t = \frac{\zeta}{n} \tag{4.4}$$

Thus a new n-component vector [Y] is created

$$Y_{j} = j * \Delta t$$
 $j = 1, ..., n$ (4.5)

IV - The physical position vector $[Z_j]$ corresponding to the vector $[Y_j]$ is obtained using the linear interpolation formula

$$Z_{j+1} = Z_K + \frac{(Z_{K+1} - Z_K) (Y_j - Y_K)}{(Y_{K+1} - Y_K)}$$
 $j=1,..., n-1$ (4.6)



or

$$Z_{j+1} = Z_K + \frac{h (Y_j - Y_K)}{N (Y_{K+1} - Y_K)}$$
 (4.7)

Thus the line is now divided into n sections of equal delay.

4.2 The main equations

Rewrite equation (3.1)

$$\frac{d}{dt} \left[V(Z,t) \pm \sqrt{\frac{L(Z)}{C(Z)}} \quad i(Z,t) \right] =$$

$$- \left[\frac{G(Z)}{C(Z)} V(Z,t) \pm \sqrt{\frac{R(Z)}{L(Z)} \cdot C(Z)} \right] \quad i(Z,t)$$
(4.8)

Let

$$W(Z,t) = V(Z,t) + \sqrt{\frac{L(Z)}{C(Z)}} i (Z,t)$$
 (4.9)

$$f(Z,t) = -\left[\frac{G(Z)}{C(Z)} V(Z,t) + \frac{R(Z)}{\sqrt{L(Z) C(Z)}} i(Z,t)\right]$$
 (4.10)

Thus one of the two equations given by (4.8) is transformed to

$$\frac{d W(Z,t)}{dt} = f(Z,t) \tag{4.11}$$

Consider the time interval [t-T, t] with $T=\Delta t$ being the time delay of each section as given by (4.4), then equation (4.11) can be written in a discrete form as



$$\Delta W = f_{avg} (Z,t) \cdot \Delta t \qquad (4.12)$$

Now, if we consider that at time t the wave W is at the $\ensuremath{\mathsf{K}}^{\mathsf{th}}$ node, then

$$\Delta W = W_{K}(t) - W_{K-1}(t - T)$$
 (4.13)

In section (3.6) it was shown that the second formulation is superior to the first formulation, so we adopt the former in approximating f(Z,t) avg so that we take

$$f(Z,t) \mid_{avg} = \frac{1}{2} [f_{K-1} (t - T) + f_{K}(t)]$$
 (4.14)

Thus we have for the forward propagating argument

$$V_{K}(t) + \rho_{K} i_{K}(t) - V_{K-1}(t - T) - \rho_{K-1} i_{K-1}(t - T) =$$

$$-\frac{\Delta t}{2} \left[\frac{G_K}{C_K} V_K(t) + \sqrt{\frac{R_K}{L_K C_K}} i_K(t) + \right]$$

$$\frac{G_{K-1}}{G_{K-1}} V_{K-1} (t - T) + \frac{R_{K-1}}{G_{K-1}} i_{K-1} (t - T)]$$
 (4.15)

or

$$V_{K}(t) \left[1 + \frac{\Delta t}{2} \frac{G_{K}}{C_{K}}\right] + \left[\rho_{K} + \frac{\Delta t}{2} \sqrt{\frac{R_{K}}{L_{K}C_{K}}}\right] i_{K}(t) =$$

$$V_{K-1}(t - T) \left[1 - \frac{\Delta t}{2} \frac{G_{K-1}}{C_{K-1}}\right] + \left[\rho_{K-1} - \frac{\Delta t}{2} \sqrt{\frac{R_{K-1}}{L_{K-1}C_{k-1}}}\right] i_{K-1}(t - T)$$

$$K=2, \dots, N \quad (4.16)$$

Similarly for the backward propagating wave



$$V_{K-1}(t) \left[1 + \frac{\Delta t}{2} \frac{G_{K-1}}{C_{K-1}}\right] - \left[\rho_{K-1} + \frac{\Delta t}{2} \sqrt{\frac{R_{K-1}}{L_{K-1}}}\right] i_{K-1}(t) =$$

$$V_{K}(t-T) \left[1-\frac{\Delta t}{2}\frac{G_{K}}{C_{K}}\right] - \left[\rho_{K}-\frac{\Delta t}{2}\frac{R_{K}}{\sqrt{L_{K}C_{K}}}\right]i_{K}(t-T)$$

$$K=2,...N$$
 (4.17)

Let

$$a_{K} = \Delta t \frac{R_{K}}{\sqrt{L_{K}C_{K}}}$$
 (4.18)

$$b_{K} = \Delta t \frac{G_{K}}{C_{K}}$$
 (4.19)

Then (4.16) and (4.17) transform to

$$V_{K}(t) \left[1 + \frac{b_{K}}{2}\right] + \left[\rho_{K} + \frac{a_{K}}{2}\right] i_{K}(t) =$$

$$V_{K-1}(t-T) \left[1-\frac{b_{K-1}}{2}\right] + \left[\rho_{K-1}-\frac{a_{K-1}}{2}\right] i_{K-1}(t-T)$$
 (4.20)

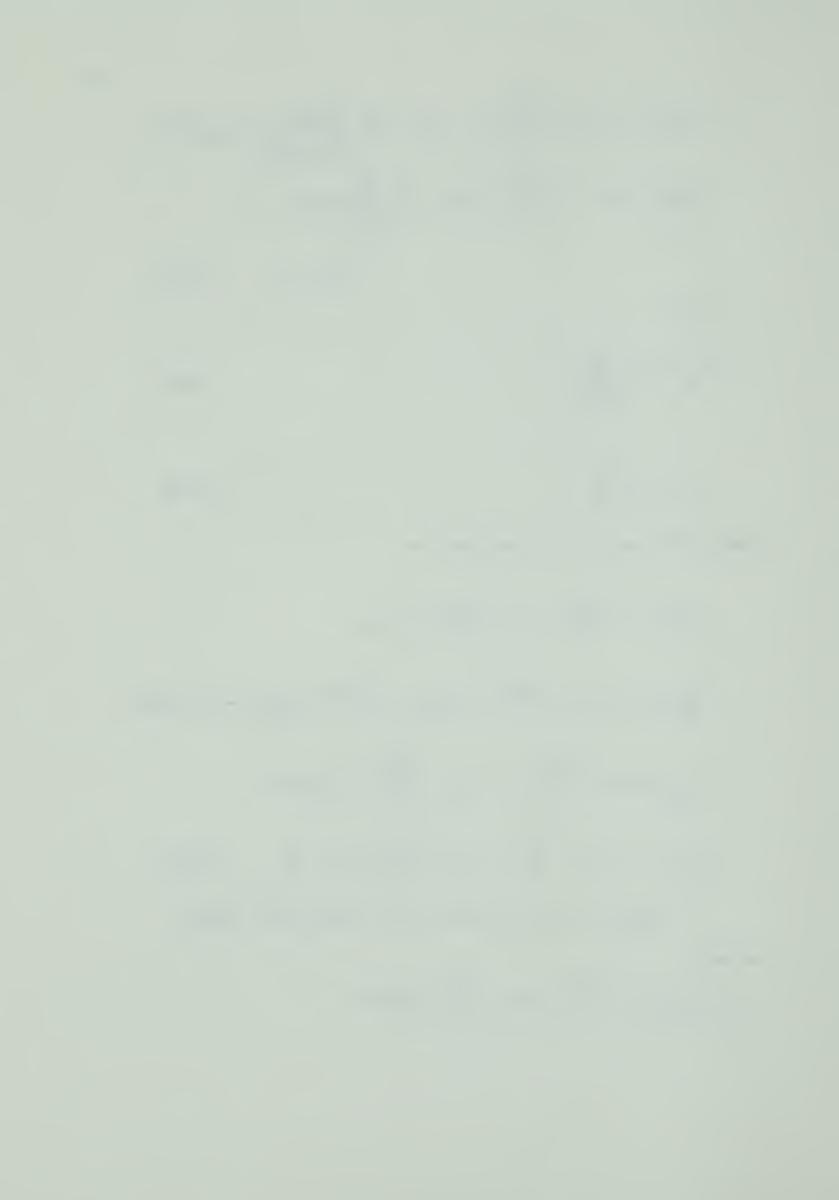
$$V_{K-1}(t) \left[1 + \frac{b_{K-1}}{2}\right] - \left[\rho_{K-1} + \frac{a_{K-1}}{2}\right] i_{K-1}(t) =$$

$$V_{K}(t-T)[1-\frac{b_{K}}{2}]-[\rho_{K}-\frac{a_{K}}{2}]i_{K}(t-T)$$
 (4.21)

Rewrite (4.21) between the K^{th} and $(K + 1)^{st}$ nodes,

we have

$$V_K(t) \left[1 + \frac{b_K}{2}\right] - \left[\rho_K + \frac{a_K}{2}\right] i_K(t) =$$



$$V_{K+1}(t-T) \left[1-\frac{b_{K+1}}{2}\right] - \left[\rho_{K+1}-\frac{a_{K+1}}{2}\right] i_{K+1}(t-T) (4.22)$$

further adding (4.20) to (4.22) and subtracting (4.22) from (4.20) we get

$$V_{K}(t) = \left[(1 - \frac{b_{K-1}}{2}) V_{K-1}(t - T) + \frac{a_{K-1}}{2} \right] V_{K+1}(t - T) + (\rho_{K-1} - \frac{a_{K-1}}{2}) i_{K-1}(t - T) - (\rho_{K+1} - \frac{a_{K+1}}{2}) i_{K+1}(t - T) \right] / \left[2 + b_{K} \right]$$

$$i_{K}(t) = \left[(1 - \frac{b_{K-1}}{2}) V_{K-1}(t - T) - \frac{a_{K-1}}{2} \right] V_{K-1}(t - T) - (1 - \frac{b_{K+1}}{2}) V_{K+1}(t - T) + (\rho_{K-1} - \frac{a_{K-1}}{2}) i_{K-1}(t - T)$$

+
$$(\rho_{K+1} - \frac{a_{K+1}}{2}) i_{K+1} (t - T)]/[2\rho_K + a_K]$$
 (4.24)



CHAPTER 5

DESCRIPTION OF COMPUTER PROGRAM

The algorithms discussed in chapters 3 and 4 were incorporated into a Fortran IV computer program. The operation and flow of the main program are described in this chapter.

A sample computer output listing is given to illustrate the operation of the program.

5.1 Computer program flow chart

The flow of the program is shown in fig. (5-1) and the program listing is given in Appendix C.

The program reads in data cards, describing the system parameters, accuracy limits and type of transmission line whose transient response is to be analyzed. The program then writes out all input parameters, type of transmission line (uniform or non-uniform) and maximum permissible relative error (in case of uniform lines).

In the case when a uniform line is analyzed, the program determines the appropriate number of sections n. If the computed value of n is less than 10, the program sets n to be equal to 10 and the main loop is entered. Also if the computed value of n is less than 1000 the main loop is entered. If the computed value of n is larger than 1000, the program writes out a message indicating that the number of sections has exceeded the maximum permissible value and the program stops. It should be noted that this limit



can be increased if desired by a minor change in the program at the expense of higher storage requirements.

If the line is nonuniform, the calculation of n is skipped and the main loop is entered. Here n is given in the data.

The first step in the main loop is to divide the line's physical length into one thousand sections of equal length. It is assumed that the line is of unit length. The vector \mathbf{Z}_K of physical length between the sending end and the \mathbf{K}^{th} node is created. Then the corresponding electrical length vector \mathbf{Y}_K is computed using a trapezoidal rule formulation.

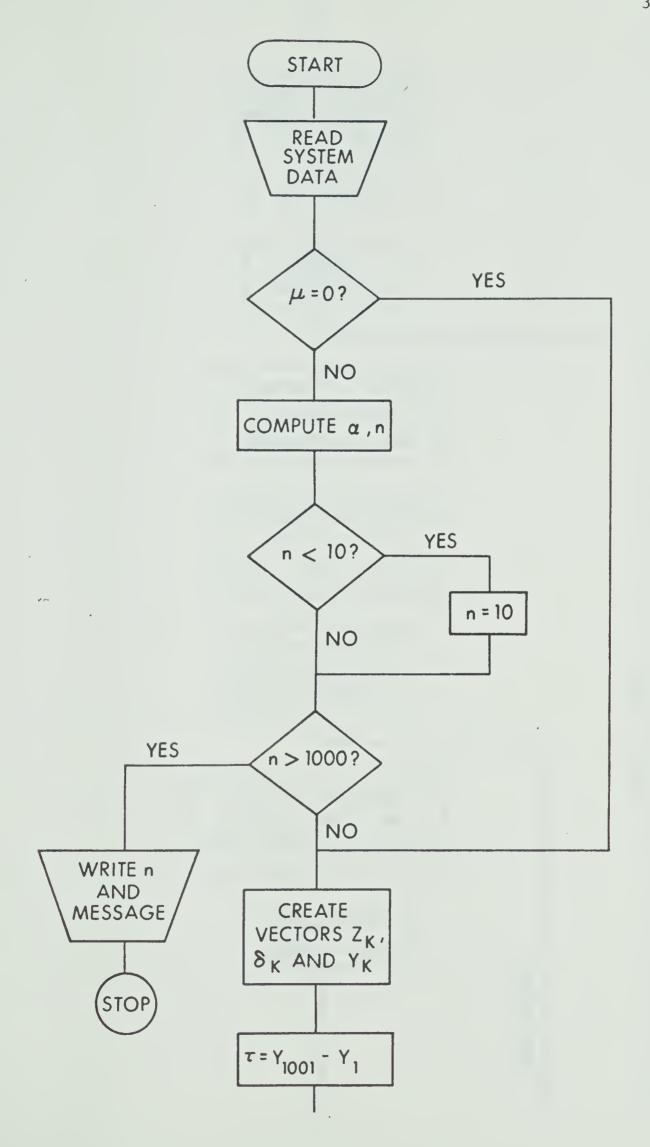
The line's electrical length ζ is then computed and the vector of normalized electrical length is created $Y_K^{(n)}$ by dividing Y_K through ζ .

The program then computes the appropriate time step Δt as the out-come of dividing the total electrical length by the number of sections n. The line is then divided into n sections of equal electrical length (delay) by creating the vector Y_j according to the rule $Y_j = \Delta t \cdot j$.

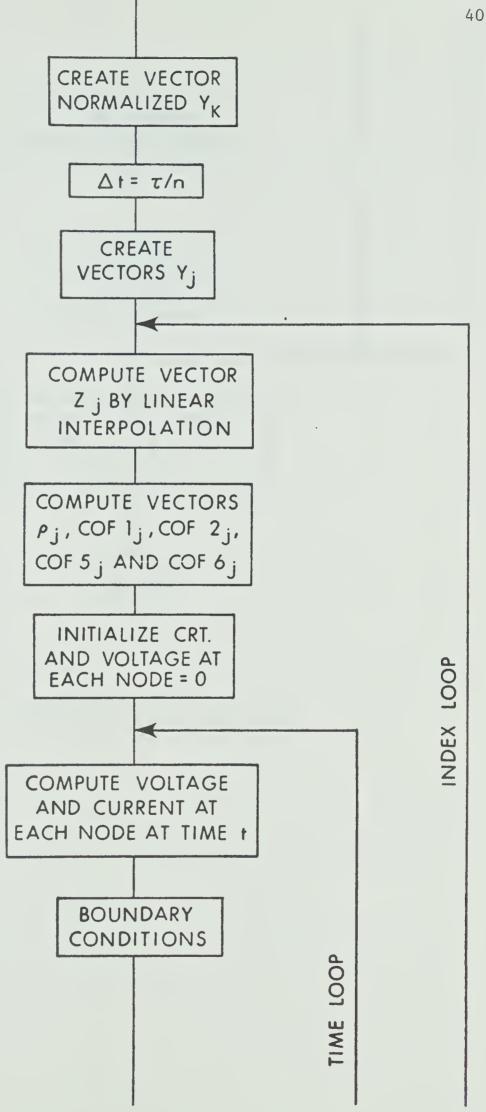
Further the physical length vector Z_j corresponding to Y_j is computed using a linear interpolation formula. With the vector Z_j created, the vectors ρ_j , Cof_{1j} , Cof_{2j} , Cof_{5j} and Cof_{6j} are computed as a prior step to applying the current and voltage relations.

The initial values of current and voltage at each node are then stored as zeros in the C and V arrays.











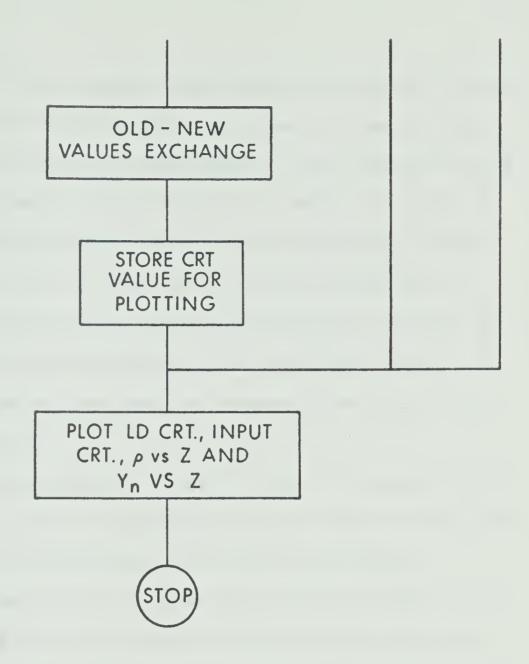


FIG. 5.1 COMPUTER PROGRAM FLOW CHART



In this program C and V denote old values of current and voltage, while CC and VV denote new values of these variables. By old and new values, it is meant values at time instants (t-n.\Delta t) and at t respectively. The next step is to compute the values of new currents and voltages at each node using old values. Then the program replaces the old values by new values to determine the currents and voltages at the next time instant and so on until the required time interval is finished. Thus the program destroys old values of voltages at every node and currents at every node except at the receiving end.

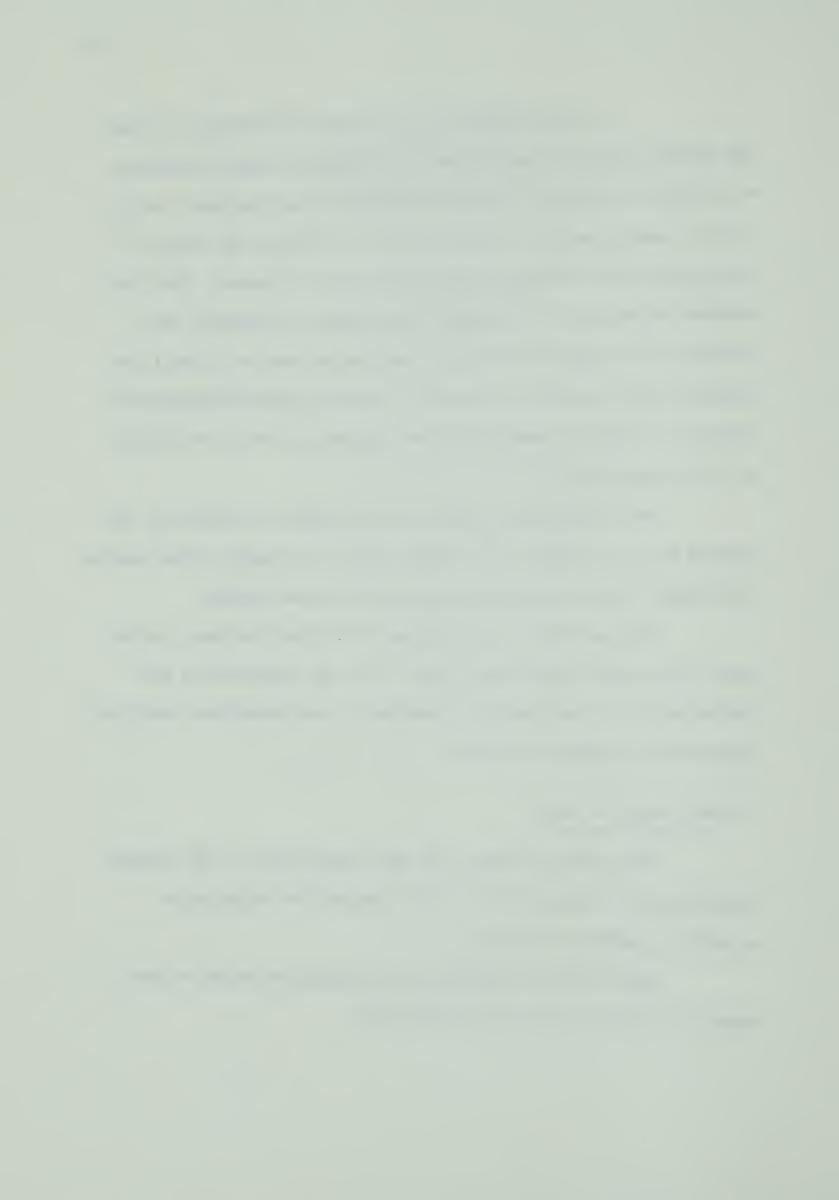
The program then writes out the values of currents at the sending end, one quarter line length, half line length, three quarter line length and receiving end at every fifth time instant.

The last step in the program is to plot the input current wave, load current wave versus time. Also the program will plot the variation of characteristic impedance ρ and normalized electrical length with the physical length.

5-2 User supplied data

The computer program user must supply data to the program describing the characteristics of the system, the calculation accuracy or number of sections.

These data are supplied to the program by means of data cards at the end of the Fortran Card deck.



The parameters and the formats on each data card are described in this section.

The first data card contains the load resistance RR and the source shunt conductance GG. The format is 2E16.8.

The second data card contains XNN and XMEW. These are the number of sections n and the relative error allowed μ . If the value of μ supplied is zero, the program takes the line as a non-uniform one. However, in this case the value of n must be entered. The format is 2E16.8.

The third data card contains RESI, RØVAL and GVAL, these values are to be entered if the line is uniform. RESI is the value of the resistance per unit length, RØVAL is the value of the characteristic impedence and GVAL is the value of the shunt conductance of the line per unit length. The format is 3E16.8.

The fourth data card contains XLO, CO, RO, GO. These are the amplitudes of inductance, capacitance, resistance and conductance per unit length functions describing the variation of these parameters with the physical length. The format is 4E16.8.

The fifth data card contains XLE1, XLE2, XLE3, AMP and SLOPE. These parameters describe the source current characteristic as a trapezoidal wave. This wave is shown in fig. 5-2 with the meaning of each of the above mentioned parameters.

It is noted that SLOPE = AMP/XLE1. The format is 5E12.4.



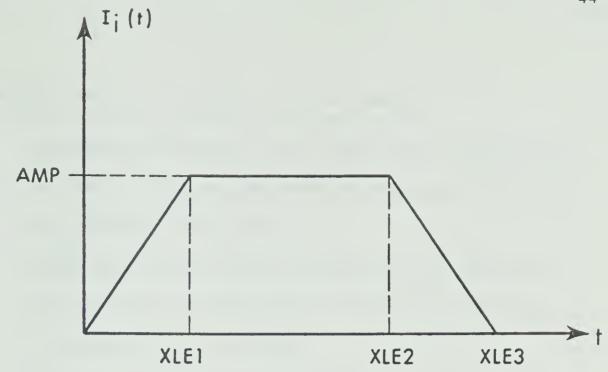


FIG. 5.2 TRAPEZOIDAL WAVE FORM

In addition to the above mentioned five data cards the following functions should be supplied.

The first function is FL(X), which describes the variation of the line's inductance with the length X. The second function is FC(X) for the capacitance. The third function is FR(X) for the resistance. The fourth function is FG(X) for the shunt conductance, and the fifth function is the source current function SOUR(X).

As an example the following functions were assumed for the above mentioned variations.

$$FL(X) = XL0* (2.0 + SIN(\pi X))$$

$$FC(X) = CO * EXP (-X)$$

$$FR(X) = RO * Log (1 + X)$$

$$FG(X) = GO$$



5-3 Numerical study of the effect of n on results

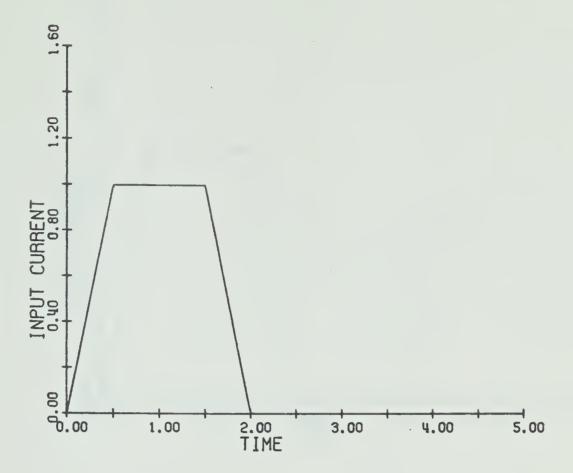
The program described in this chapter was equipped with a facility to compute and plot the transient response of the given line for three different values of n.

This is facilitated by the introduction of a three step DØ loop. The first for the given value of n say n_0 , the second for $n = 5n_0$ and the third for $n = 20n_0$.

Another trial was for n_0 , 1.5 n_0 and $2n_0$ respectively.

The computed values of currents at the locations on the line previously mentioned and plots of load current versus time are shown in figs. (5-3) and (5-4).





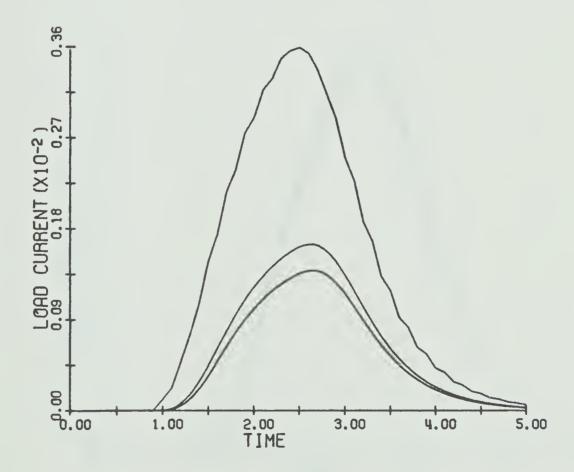
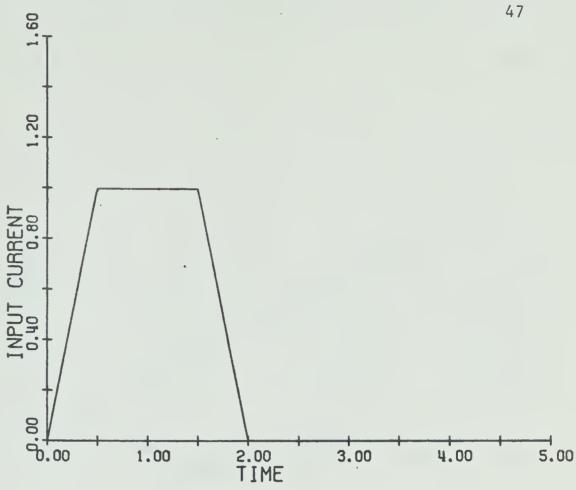


FIG. 5.3 TRANSIENT RESPONSE FOR n=10, 50 AND 200







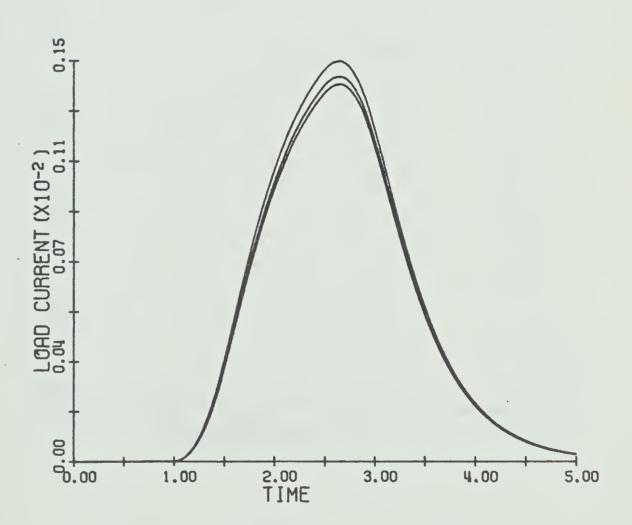
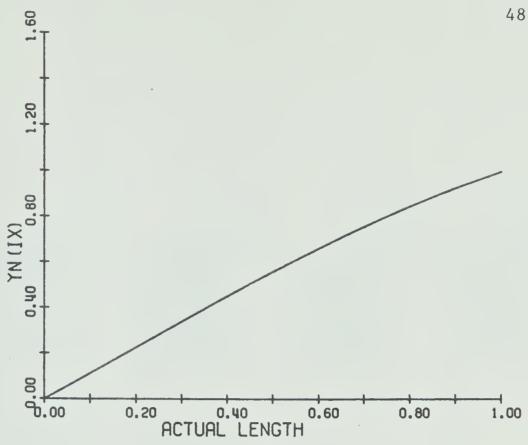
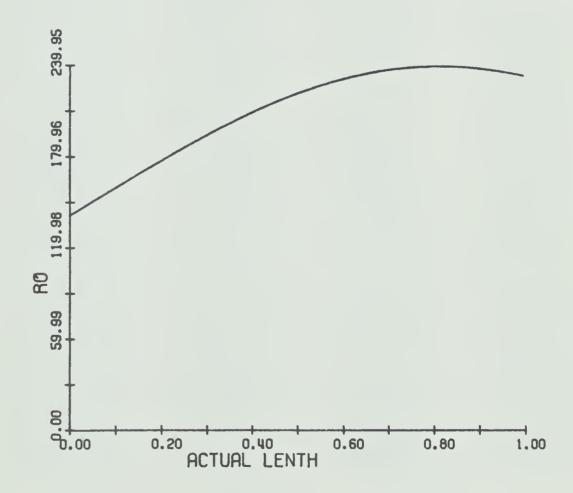
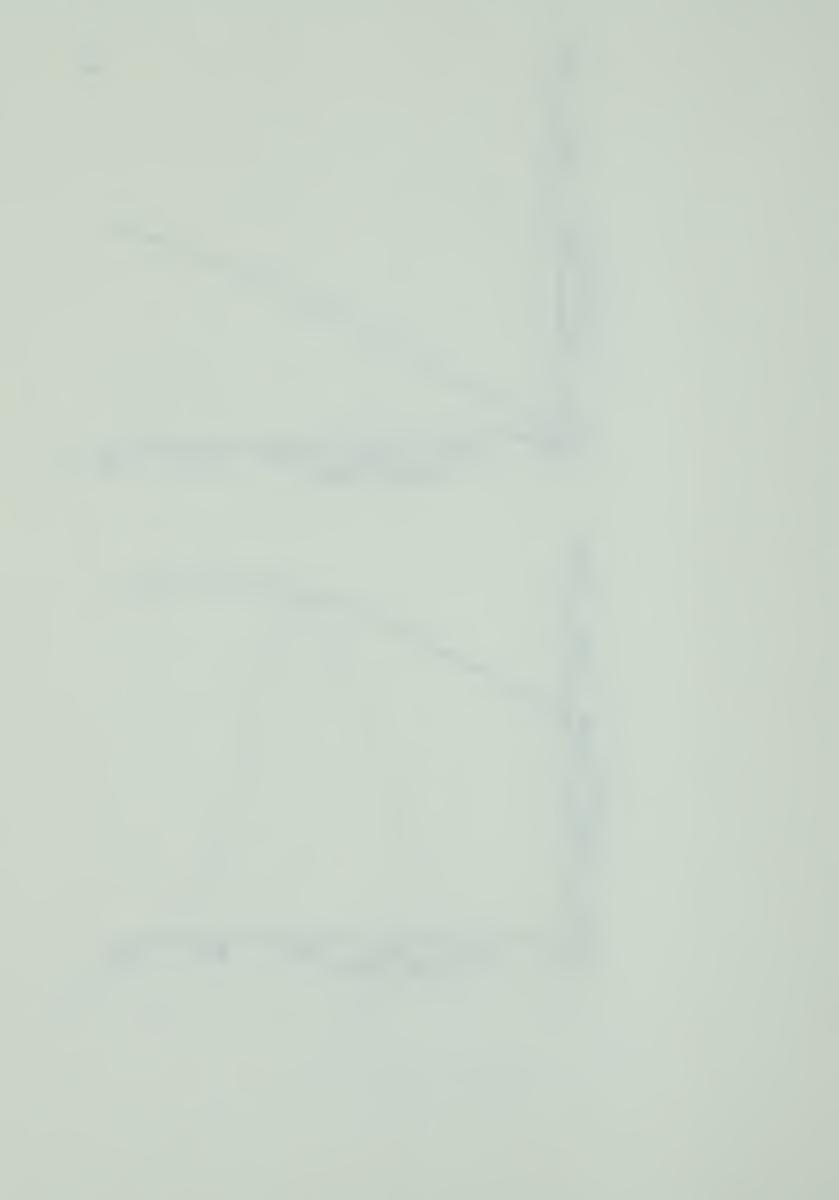


FIG. 5.4 TRANSIENT RESPONSE FOR n=100, 150 AND 200









0.800000E		0.4569398E 00	0.1179693E-01	0 • 0	0.0
0.1C00000E		0.6435866F 00	0.9511560F-01	0.1199614E-01	0.1093948E-03
0.100000E		0.6685686F 00	0.1312146E 00	0.2925690F-01	0.1458441E-02
0.2(00008F	СО	0.2170602E 00	0.1331436E 00	0.3801805F-01	0.2866412E+02
0.0		0.324856CF-01	0.5361597E-01	0.2876412E-01	0.3550411F-02
0.0		0.8078363E-02	0.1906353E-01	0.1266391E-01	0.2467589E-02
0.0		0.2882114E-02	0.5782910E-02	0.4243717F-02	0.1174732F-02
0.0		0.8772933F-03	0.2175869E-02	0.1643409E-02	0.4155547E-03
0.0		0.3304961F-03	0.6828641E-03	0.5241956E+03	0.1636188F-03
0.0		0.1037963F-03	0.2597007E-03	0.2003469E-03	0.5267253E-04
0.199999F		0.0	0.0	0.0	0.0
0.4(COCOOE		0.0	0.0	0.0	0.0
0.599999F		0.3080618F-01	0.0	0.0	0.0
0.8(00000E		0.7996100E-01	0.0	0.0	0.0
0.959999F		0.1365302E 00	0.6195493F-03	0.0	0 • 0
0.1000000E		0.1985610F 0C	0.5232073F-02	0 • 0	0 • 0
0.1CC0000E		0.2641563E 00	0.1269520E-01	0.0	0.0
0.1000000F		0.3008220F 00	0.2226996E-01	0.2227691E-03	0 • 0
0.1C00000E		0.3229754F CC	0.3367817F-01	0.8659367F-03	0.0
0.1000000F		0.3379491F 00	0.4582062E-01	0.1894894E-02	0 • 1514323E-05
0.1CCOOOOF		0.3494176E 00	0.5492588F-01	0.3316215F-02	0.2057997E-04
0 • 1 C 0 0 0 0 0 F	-	0.3576394E 00	0.6248513F-01	0.5052412E-02	0.7116888E-04
0.1000000E		0.3641481E 00	0.6832713E-01	0.6860722E-02	0.1561420E-03
0.1(C0000E		0.3689565F 00	0.7317370E-01	0.8544430F-02	0.2795046E-03
0.1C00000F		0.3728333E CC	0.7692069E-01	0.1000185E-01	0.4355770E-03
0.8C00011E		0.3757478E 00	0.8003211E-01	0.1129845F-01	0.6054251E-03
0.6C0C004F		0.3781241E 0C	0.8244067E-01	0.1237013E-01	0.7764511E-03
0.4000015F		0.3491244E 00	0.8444303E-01	0.1329944F-01	0.9337922E-03
0.2C00008F		0.3014531E 00	0.8599538E-01	0.1405025E-01	0.1079261E-02
0.4000092E-	- 01	0.2460197E 00	0.8666772E-01	0.1469234E-01	0.1203923F-02
0.0		0.1849270F 00	0.8305788E-01	0.1520436F-01	0.1314523E-02
0.0		0.1200532E 0C	0.7643008E-01	0.1563859E+01	0.1405859E-02
0.0		0.8398533E-01	0.6750405E-01	0.1575930F-01	0.1485075F-02
0.0		0.6229385E-01	0.5663674E-01	0.1540590E-01	0.1549115E-02
0.0		0.4770354E-01	0.4491479F-01	0.1460493F+01	0.1602399E-02
0.0		0.3653200E-01	0.3616019E-01	0.1337530F-01	0.1627061F-02
0.0		0.2855720E-01	0.2887370F-01 0.2325925E-01	0.1177942E-01	0.1613578E-02 0.1557969E-02
0.0		0.2223997E-01		0.1010722E-01	
0.0		0.1759101E-01 0.1383808E-01	0.1858965E-01 0.1499033E-01	0.8522157F-02 0.7147375E-02	0.1459330F-02 0.1322827F+02
0.0		0.1102665E-01	0.1199392F-01	0.5915325F-02	0.1169388E=02
0.0		0.8730456E-02	0.9681161E-02	0.4897650F-02	0.1011231E-02
0.0		0.6990764E-02	0.7753406E-02	0.4010502E-02	0.8646892E-03
0.0		0.5558982F-02	0.6263349E-02	0.3294913E-02	0.7276691F-03
0.0		0.4466049E-02	0.5019918E-02	0.2680305E-02	0.6100861E-03
0.0		0.3561905E-02	0.4057627E-02	0.2191249E-02	0.5050148F-03
0.0		0.2868184E-02	0.3253882E-02	0.1774923E-02	0.4183054F-03
0.0		0.2292232F-02	0.2631263E-02	0.1446396F-02	0.3427006E-03
0.0		0.1848745F-02	0.2110868E-02	0.1168286F-02	0.2816790E-03
0.0		0.1479618E-02	0.1707478E-02	0.9500002F-03	0.2292354E-03
0.0		0.1194675F-02	0.1370152F-02	0.7658843F-03	0.1874745F-03
0.0		0.9570925F-03	0.1108545E-02	0.6218820E-03	0.1519041E-03
0.0		0.7733721E-03	0.8897078E-03	0.5007146E-03	0.1238189E-03
0.0		0.6200001F-03	0.7199359E-03	0.4061693E-03	0.1000340E-03
0.0		0.5012546E-03	0.5778880E-03	0.3267457E-03	0.8135765E-04
		V 30 12 2 4 0 12 0 3			3,11,0,7,0,0,0,0,0,0



0.0	0.4020391E-03	0.00744075.07	0.04.00000.03	0 65600305 04
0.0	0.3251587E-03	0.4676627E-03	0.2648709E-03	0.6560038E-04
0.0	0.2608846E-03	0.3754226E-03	0.2129504E-03	0.5327264F-04
0.0	0.2110510F-03	0.3038356E-03	0.1725454E-03	0.4289765E-04
0 • 0		0.2439232E-03	0 • 1386658E = 03	0.3480361F-04
0.ECC0000E-01	0.1693717E-03	0.1974214E-03	0.1123202E-03	0.2799767E-04
	0 • 0	0.0	0.0	0 • 0
0.9549996F-01	0.0	0.0	0.0	0.0
0.1500000F 00	0 • 0	0.0	0.0	0 • 0
0.159999F CO	0 • 0	0.0	0.0	0.0
0.2500000E 00	0.0	0 • 0	0.0	0.0
0.3CC0C00E 00	0.0	0.0	0.0	0.0
0.349999E 00	0.0	0.0	0 •0	0.0
0.4(00000F CO	0.0	0.0	0.0	0.0
0.4499999E 00	0 • 0	0.0	0.0	0.0
0.500000F 00	0.1321034F-n2	0.0	0.0	0.0
0.5500000E 00	0.8377887E-02	0.0	0.0	0.0
0.590999F 00	0.1633499E-01	0.0	0.0	0 • 0
0.6500000F 00	0.2502728F-01	9.0	0.0	0.0
0.6596998 00	0.3442798E-01	0 • 0	0 • 0	0 • 0
0.75CC000E CO	0.4441303E-01	0.0	0.0	0.0
0 • 8000000E 00	0.5496659F-01	0 • 0	0.0	0.0
0.849999F 00	0.6599307F-01	0.0	0.0	0.0
0.9000000000000	0.77483536-01	0.0	0 • 0	0.0
0.9449999F 00	0.89362386-01	0.0	0.0	0.0
0.9899999F CO	0.1016248F 00	0 • 1083283E=03	0.0	0.0
0.1000000 01	0.1142105F CC	0.7417507E-03		
0 • 1 CC 0 0 0 0 E 0 1	0.1271170F CC		0.0	2.0
0.1CC0000E C1	0.1402954E 00	0.1567815E-02	0.0	2 • 0
		0.25032748-02	0.0	0 • 0
0 • 1 0 0 0 0 0 0 1	0.1537443F 0C	0.3733363F-02	0 • 0	0 • 0
0.1(00000F 01	0.1674239E 00	0.5056720F-02	0.0	0 • 0
0.1CC0000F 01	0.1813340E (C	0.6537829E-02	0 • 0	0.0
0 • 1 COOOUDE 01	0.19544116 (0	0.8157190E-02	0 • 0	0.0
0.1(000000 01	0.2097450F 00	0.9918757F-02	0.0	9 • 0
0 • 1 0 0 0 0 0 F C 1	0.2242184E CC	0.11804776-01	0 • 0	0 • 0
0 • 1 C C O 2 O O F C 1	0.2375407E 00	0.1381867E-01	0.5070567F-05	0 • 0
0.1CCC000E 01	0.2452039F 00	0 • 1594450E-01	0.3864350F-04	0.0
0.100000F C1	0.2522522E 00	0.1818518E-01	0.90812865-04	0 • 0
0.1000000F 01	0.2585679F 00	0.2052644E-01	0.1610400F-03	0.0
0.1C00000F 01	0.2644120E (C	0.2297085E-01	0.2515251E-03	0.0
0.10000000 01	0.2696837F 00	0.2550543E-01	0.3611504E-03	0 • 0
0.100000F 01	0.2745857E 00	0.2813262F-01	0.4917916E-03	0.0
0.1(00000F 01	0.2790315E CC	0.3084071E-01	0.6419457E-03	0.0
0 • 1 C 0 0 0 0 0 F 0 1	0.2831826E 00	0.3363174E-01	0.8132237F-03	0.0
0.1CC0000E C1	0.2869637E 00	0.3649534E-01	0.1003881E-02	0.0
0.1000000E 01	0.29050615 00	0.3932513E-01	0.1215307E-02	0.1607165E-06
0.1C00000f 01	0.2937447F CC	0.41679930-01	0.1445627F-02	0.1401719E-05
0.10000000001	0.29678708 00	0.4393852F-01	0.1696036E-02	0.3724990E-05
0.1C000C0F 01	0.2995774F CC	0.46057665-01	0.1964609E-02	0.7217894E-05
0.1C00000F C1	0.3022044F CO	0.4808897F-01	0.2252384E-02	0.1213886F-04
0.1000000 01	0.30461975 00	0.4999368E-01	0.2557433F-02	0.1850646E-04
0.10000000 01	0.3069580E 00	0.51F1837E-01	0.2880663E-02	0.26558296-04
0.10000000 01	0.30899765 00	0.5352866F=01	0 • 3220182E-02	0 • 3625485E=04
0.10000001 01	0.4109815F CC	0.55166656-01	0.3576786E-02	0.4781148E-04
0.1000000000000000000000000000000000000	0.31281288 00	0.5670145F-01	0.3948644E-02	0.6114223F=04
	0.31454616 00	0.58171166-01	0.4331373F-02	0.7644035E-04
0.1000000001		0.5754764E=01		0.9358450E-04
0.1C00000F 01	0.3161484F 00		0.4698392F-02	
0.10000000001	0.3176667E 00	0.6065564E=01	0.5064506F-02	0 · 1127468F - 03
0.1000000F 01	0.3190724F 00	0.6210001F-01	0.54227456-02	0.1337900F-03
0.10000001 01	0.3204058F 00	0.63231836-01	0.5777583F-02	0 • 1568 373E = 03
0.10000000001	9.3210416F 00	0.64388335-01	0.61222656-02	0.1817536F-03



0.100000E 0	1 0.32281498	00 0.	6544775E-01	0.6461833E-02	0.2086637E-03
0.1000000E 0	1 0.32390436	00 0.	6643981E-01	0.6789856E-02	0.2373929E-03
0.100000E 0	1 0.32493896		67 38943E-01	0.7111639E-02	0.2680614E-03
0.1000000E 0	1 0.32590008		6827867E-01	0.7421110E-02	0.3004870E-03
0.100000E 0	1 0.32681408		6913006E-01	0.7723689E-02	0.3346147E-03
0.9500008E 0			6992698E-01	0.8013632E-02	0.3692608E-03
0,9000015E 0			7068992E-01	0.8296363E-02	0.4047656E-03
0.8500004E 0		= -	7140434E-01		
0.8000011E 0				0.8566506E-02	0.4405011E-03
0.75C0000F 0			7208836E-01	0.8829344F-02	0.4767065E-03
0.70000AF 0			7272875E-01	0.9079855E-02	0.5127268F-03
	_		7334179E-01	0.9323120E-02	0.5488985E-03
0.6500015E 0			7391602E-01	0.9554554E-02	0.5845579E-03
0.6(CC004E 0			7446557E-01	0.9778913E-02	0.6201160E-03
0.550C011F C			7498032E-01	0.9992000E-02	0.6549151E-03
0.500000F 0		00 0.	7547313E-01	0.1019832E-01	0.6894181E-03
0.4500008F 0	0 0.32551596	00 0.	7593453E-01	0.1039394E-01	0.7229832E-03
0.4000015F C	0 0.3180048F	00 0.	7637632E-01	0.1058311E-01	0.7561084E-03
0.3500004F 0	0 0.3097282F	0.	7679015E-01	0.1076230E-01	0.7881739E-03
0.3000011E C	0 0.3007243E	00 0.	7718629E-01	0.1093538E-01	0.8196994E-03
0.2500000F 0	0 0.29110976	00 0.	7755738E-01	0.1109913E-01	0.8500898E-03
0.200000BE 0	0 0.28090936	00 0.	7791263F-01	0.1125716E-01	0.8798703E-03
0.1500015F 0	0 0.27021296		7824528E-01	0.1140654F-01	0.9084775E-03
0.1000004E 0			7856405E-01	0.1155058F-01	0.9364360E-03
0.5000114F-0			7886249E-01	0.1168662E-01	0.9632134E-03
0.1000023E-0			7903999E-01		0.9893202E-03
0.0	0.22314836			0.1181771E-01	
0.0			7867420E-01	0.1194140E-01	0.1014262E-02
	0.21049306		7810438E-01	0.1206053E-01	0.1038530E-02
0.0	0.19754916		7734889E-01	0.1217287E-01	0.1061665E-02
0.0	0.18432436		7640886E-01	0.1228101E-01	0.1084134E-02
0.0	0.17085426		7530075E-01	0.1238292E-01	0.1105516E-02
0.0	0.15714458		7402593E-01	0.1248096F-01	0.1126248E-02
0.0	0.14322506		7259965F-01	0.1257330E-01	0.1145945E-02
0.0	0 • i 290 998E	00 0.	7102311E-01	0.1266214E-01	0.1165019E-02
0 • 0	0.11479418	00 0.	6931043E-01	0.1274575E-01	0.1183118E-02
0.0	0.10163208	00 0.	6746250E-01	0.1282105E-01	0.1200620E-02
0.0	0.94118428	-01 0.	6549227E-01	0.1286312E-01	0.1217203E-02
0.0	0.87213646	-01 0.	6340033E-01	0.1288365E-01	0.1233228E-02
0.0	0.81031866	-01 0.	6119857E-01	0.1288179E-01	0.1248393E-02
0.0	0.75315898	-01 0.	5888793E-01	0 • 12857 02E-01	0.1263033E-02
0.0	0.7016420E	-01 0.	5647861E-01	0.1280914E-01	0.1276876E-02
0.0	0.6537706E	-01 0.	5397129E-01	0.1273783E-01	0.1290228E-02
0.0	0.61038858	-01 0.	5137560E-01	0.1264345E-01	0.1302843E-02
0.0	0.5699090E	-01 0.	4869222F-01	0.1252573E-01	0.1315002E-02
0.0	0.53305726	-01 0.	4592937E-01	0.1238539E-01	0.1326479E-02
0.0	0.4985565F		4319631E-01	0.1222230E-01	0.1337375E-02
0.0	0.4670316E		4093200F-01	0.1203736E-01	0.1346559E-02
0.0	0.43743468		3876021E-01	0.1183052E-01	0.1354278E-02
0.0	0.41030646		3672235E-01	0.1160285E-01	0.1360251E-02
	0.41030646 0.3847763F		3476894F-01	0.1135435E-01	0.1364438E-02
0.0					0.1366650E-02
0.0	0.36131508		3293722E+01	0.1108619F-01	
0.0	0.3391917E		3118228E-01	0.1079832F-01	0.1366854E-02
0.0	0.31681706		2953749F-01	0.1049203F-01	0.1364928E-02
0.0	0.29957136		2796216E-01	0.1016730E-01	0.1360847E-02
0.0	0.2818130F		2648618E-01	0.9825367E-02	0.1354550E-02
0.0	0.2650142E		2507285E-01	0.9471346E-02	0.1346012F-02
0.0	0.24948916		2374899E-01	0.9131275E-02	0.1335230E-02
0.0	0.2347636E	-01 0.	2248155F-01	0.8790996E-02	0.1322179E-02
0.0	0.2211738E	-01 0.	2129447E-01	0.8457012E-02	0.1306893E-02
0.0	0.2082679E	-01 0.	2015810E-01	0.8125447E-02	0.1289356E-02
0.0	0.1963087E	-01 0.	1909391E-01	0.7802580F-02	0.1269632E-02
7.0	0.18495726	-01 0.	1807523E-01	0.7483952E-02	0.1247705E-02



0.0	0 14/12/55			
	0.1/44265E-01	0.1712132E-01	U. 7175572E-02	U.1227002F-UL
0.0	0.16442248-01	U.1620824E-01	0.0672620E-02	0.119749ZE=JZ
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0.0	0.2735618E-03	0.2978207E-03	0 • 1 585595E - 03	0.3949087F-04
0.0	0.2452426E-03	0.2676852F-03	0.1422625E-03	0.3554061E+114
0.0	0.2203809F-03	0.2400350E-03	0.1279330E-03	0.31908196-04
0.0	0.19757676-03	0.2157496E-03	0.1147736F-03	0.28710338 = 1 4
0.0	0.1774551F-03	0.1934659E-03	0.1032049F-03	0.25770906- 4
0.6666666E-01	0.0	0.0	0 • 0	0 • 0
0.133333B CC	0.0	0.0	0.0	0.0
0.199999F CO	0.0	0.0	0.0	0.0
0.266667F CO	9.0	0.0	0.0	0.1
0.2333333E CC	0.0	0 • 0	0.0	0.0
C.4CGOEGGE CO		0.0	0.0	0.0
	0.0			
0.4666666 30	0.77.70.20.00	0.0	0.0	0.0
0.50 BBB 60	0.77170321-07	0.0	0.0	0.7
0.5000001 00	0.15812176-01	0.0	0 • 0	0.0
n.creeted co	0.31208916-01).0	0.0	0.0
0.7233336 (0	0.44813211-01	0.0	0 • 0	0.0
0 *4CC0000E 00	0.5937929F-01	0 • 0	0.0	0.0
v.teroteef co	0.74df.205t-01	() • (0.0	0.0
0.433336 00	0.01047466-01	0.0	0.0	0.0
0.48/6/6/6/6	0.10796266 00	0.15069815-03	0 • 0	0.0



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0 - 1 COOOOOE 01
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                                            0.2325776E-02
                                                                                       0.0
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0 - 1 C 0 0 0 0 0 F
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                      0.1619442E
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0.1C00000E 01
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                                            0.5726505F-02
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                                                                                       0.0
0.1C00000E
            01
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                                            0.7832032E-02
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0 - 1 COO O O O E O 1
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                                                                                       0.0
0.100000E 01
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                                            0.1278390E-01
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                                                                                       0.0
                      0.2521293F
0.1 C00000F 01
                                  00
                                            0.1559598F-01
                                                                 0.3544339E-04
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0 - 1 COOOOOE 01
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0 - 1 C C O O O O F 0 1
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0 - 1 CC 0 0 0 0 E 01
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0.1999999E 00	/O • O	0.0	0.0	0.0
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0.0	0.3246865E-03	0.3476513E-03		
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0.0	0.2913983E-03	0.3120967E-03	0.1635580E-03	0.3994025E-04
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		0 10303335 03	0.10004545	0 0/30/13/ 61
0.0	0.1791080E-03	0.1920232E-03	0.1008655E=03	0.2470617E-04
0.0		0.1920232E-03 0.23995053E 03	0.1008655E-03	0.2470617E-04



CHAPTER 6

CONCLUSION

A method for the evaluation of the transient response of a nonuniform lossy transmission line has been presented in this thesis. The pair of partial differential equations which describes the voltage and current relations on the line is transformed into a pair of ordinary differential equations on two characteristic curves using the method of characteristics.

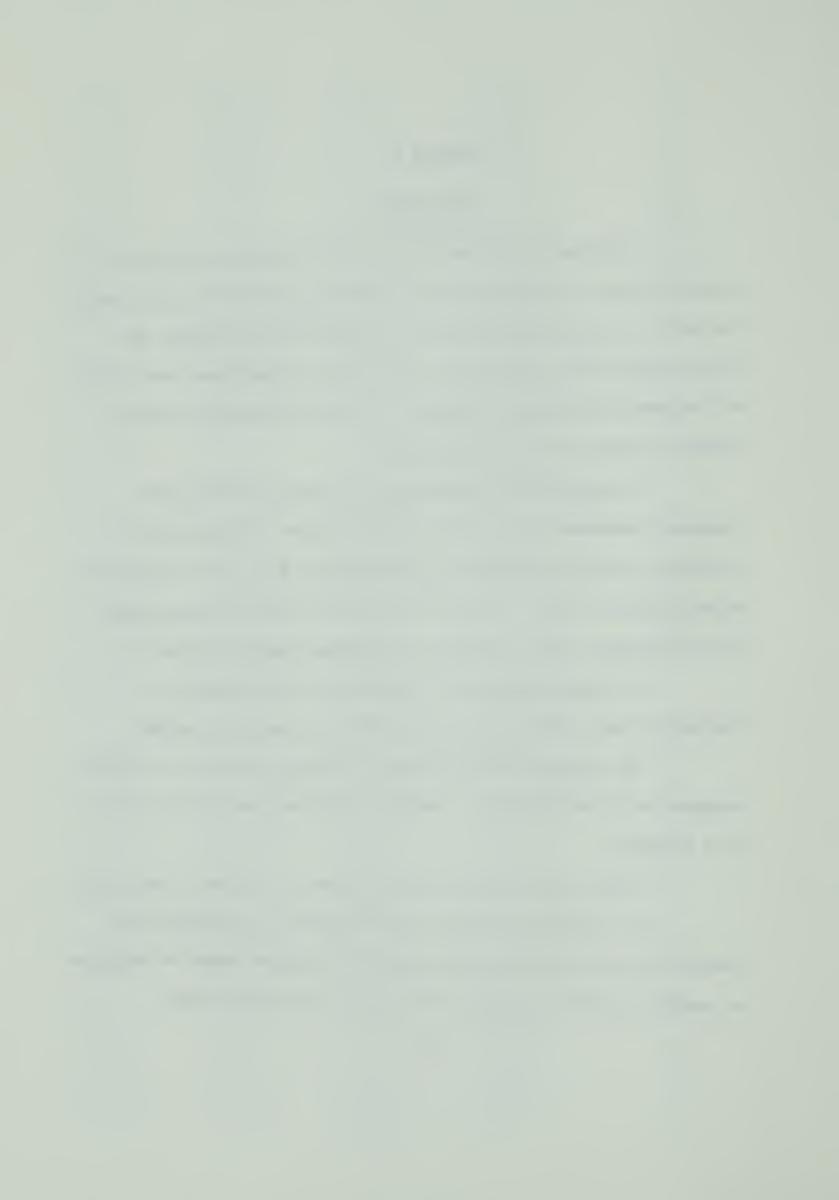
A stepped line approximation is used to analyze the transient response of the given nonuniform line. The concept of electrical length is employed in dividing the line into a number of equal delay sections. The set of difference equations describing the set stepped line is suitable for digital computer solution.

The main advantage of this method is that numerical techniques such as the Runge-Kutta method are entirely avoided.

The computational procedure involving the use of a digital computer is illustrated for a specific distributions of L(x), C(x), r(x) and g(x).

There is still much work which has to be done in this area.

For instance, it will be worth while to investigate the possibility of obtaining an estimate of the proper number of sections to achieve a given accuracy in the case of nonuniform lines.



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APPENDIX A

BASIC EQUATIONS FOR A UNIFORM LOSSLESS LINE

Consider a uniform lossless transmission line of length l as shown in figure (A.1). The following two equations hold true for the voltage and current values w.r.t. the position on the line x and time.

$$\frac{\partial^2 V(x,s)}{\partial x^2} = s^2 L C V (x,s)$$
 (A-1)

$$\frac{\partial^2 I(x,s)}{\partial x^2} = s^2 L C I (x,s)$$
 (A-2)

Where L and C are the inductance and capacitance per unit length of the line respectively, and s being the Laplace operator.

The solution of A-1 and A-2 takes the form

$$V(x,s) = F(s) e^{\frac{-sx}{v}} + G(s) e^{\frac{sx}{v}}$$
(A-3)

$$\rho I(x,s) = F(s) e^{\frac{-sx}{V}} - G(s) e^{\frac{sx}{V}}$$
(A-4)

where

$$v = \frac{1}{(LC)^{\frac{1}{2}}} \tag{A-5}$$

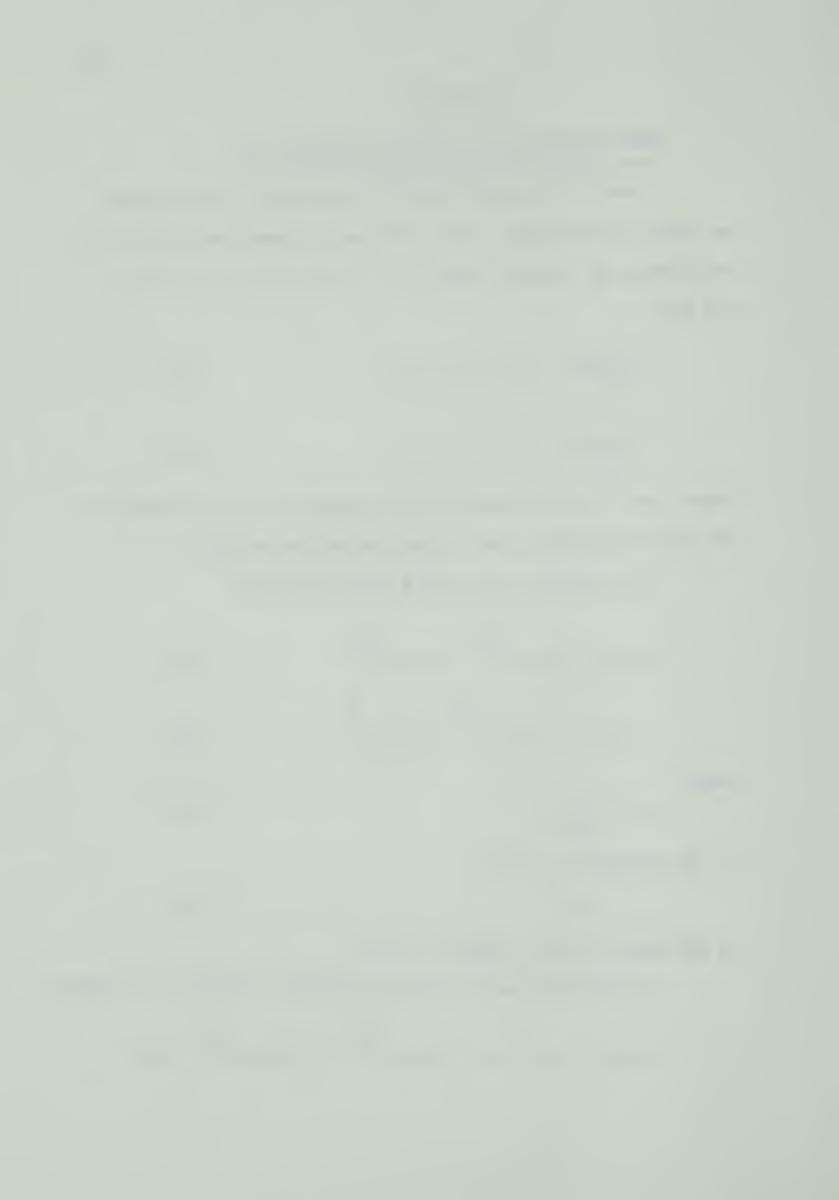
is the propagation velocity.

$$\rho = (L/C)^{\frac{1}{2}} \tag{A-6}$$

is the characteristic impedance of line.

Applying A-3 and A-4 at the boundaries, x=0 and x=1 we obtain:

$$V(1,s) + \rho I (1,s) = V(0,s) e^{\frac{-sL}{V}} + \rho I (0,s) e^{\frac{-sL}{V}}$$
 (A-7)



$$V(0,s) - \rho I (0,s) = V(1,s) e^{\frac{-sL}{V}} - \rho I(1,s) e^{\frac{-sL}{V}}$$
 (A-8)

Letting

$$T = \frac{1}{v}$$

Then we get in the time domain

$$V(1,t) + \rho i(1,t) = V(0, t-T) + \rho i(0, t-T)$$
 (A-9)

$$V(0,t) - \rho i(0,t) = V(1, t-T) - \rho i(1, t-T)$$
 (A-10)

It should be noted that the classical transmission line analysis adopted here assumes the following:

- 1. The transverse electric field components in the conductor are negligible compared with the axial. This is equivalent to the assumption that displacement currents in the conductor are negligible compared to conduction currents. This is stated mathematically as $\frac{\sigma}{m\epsilon} >> 1$.
- 2. The axial electric field components in the dielectric are small compared with the transverse. This is equivalent to the assumption that the characteristic impedance of the dielectric is much greater than the skin effect surface resistivity of the conductor or $\frac{R}{n}$ <<1.

These inequalities are nearly always satisfied by the materials of common transmission lines, but if they are not, one must examine critically any results predicted by the usual transmission line equations. Also, the longer the duration of the input current function, the better these inequalities are met.



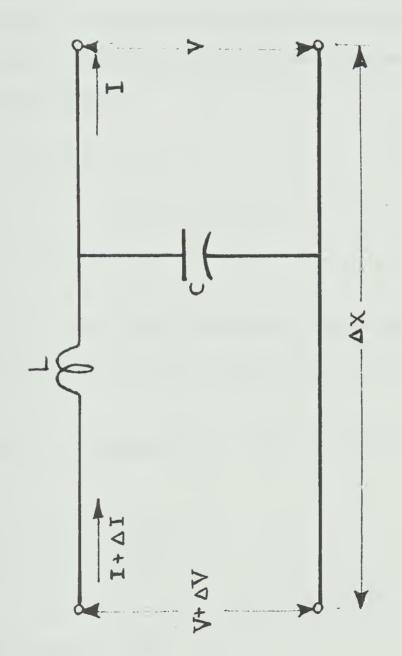


Fig. A-1 A Uniform Losseless Transmission Line



APPENDIX B

DERIVATION OF EQUATION (3-50)

Consider an incremental length Δx at a distance x from the sending end of a uniform lossy transmission line with length h. This incremental length is shown in fig. B-1, then we can write the following equations:

$$-\frac{\partial V}{\partial x} = (sL + R) I$$
 (B-1)

$$-\frac{\partial I}{\partial x} = (sC + G) V$$
 (B-2)

where R, L, C and G are resistance, inductance, capacitance and conductance per unit length respectively, s is the Laplace operator.

Partial differentiation of (B-1) and (B-2) w.r.t. x yields:

$$-\frac{\partial^2 V}{\partial x^2} = (sL + R) \frac{\partial I}{\partial x}$$
 (B-3)

$$-\frac{\partial^2 I}{\partial x^2} = (sC + G) \frac{\partial V}{\partial x}$$
 (B-4)

Substituting for $\frac{\partial I}{\partial x}$ and $\frac{\partial V}{\partial x}$ in (B-3) and (B-4) from (B-2) and (B-1)



respectively yields

$$\frac{\partial^2 V}{\partial v^2} = (sL + R) (sC + G) V$$
 (B-5)

$$\frac{\partial^2 I}{\partial x^2} = (sL + R) (sC + G) I$$
 (B-6)

Let

$$\gamma^2$$
 (s) = (sL + R) (sC + G) (B-7)

Then (B-5) and (B-6) can be rewritten as:

$$\frac{\partial^2 V}{\partial x^2} = \gamma^2 \quad (s) \quad V \tag{B-8}$$

$$\frac{\partial^2 I}{\partial x^2} = \gamma^2 \quad (s) \quad I \tag{B-9}$$

The solution to (B-8) is

$$V(x,s) = F(s) e^{-\gamma x} + H(s) e^{\gamma x}$$
(B-10)

where F(s) and H(s) are functions to be determined such that boundary conditions along the line are satisfied. Using (B-10) and (B-1) we get for I(x,s):

$$I(x,s) = \frac{\gamma(s)}{sL+R} [F(s) e^{-\gamma x} - H(s) e^{\gamma x}]$$
 (B-11)

In (B-10) and (B-11), forward propagating components of voltage and current are the ones having negative exponents, and



backward propagating components are those with positive exponents.

Thus if suffix f denotes forward components then we have

$$V_f(x,s) = F(s) e^{-\gamma x}$$
 (B-12)

$$I_{f}(x,s) = \frac{\gamma(s)}{sL+R} \quad F(s) \quad e^{-\gamma x}$$
 (B-13)

now at x = 0 we have

$$V_f(0,s) = F(s)$$
 (B-14)

$$I_f(0,s) = \frac{\gamma(s)}{sL+R} F(s)$$
 (B-15)

Thus (B-12) and (B-13) rewritten as

$$V_f(x,s) = V_f(0,s) e^{-\gamma x}$$
 (B-16)

$$I_f(x,s) = I_f(0,s) e^{-\gamma x}$$
 (B-17)

Now by definition of W(x,s) we have

$$W(x,s) = V_f(x,s) + \rho I_f(x,s)$$
 (B-18)

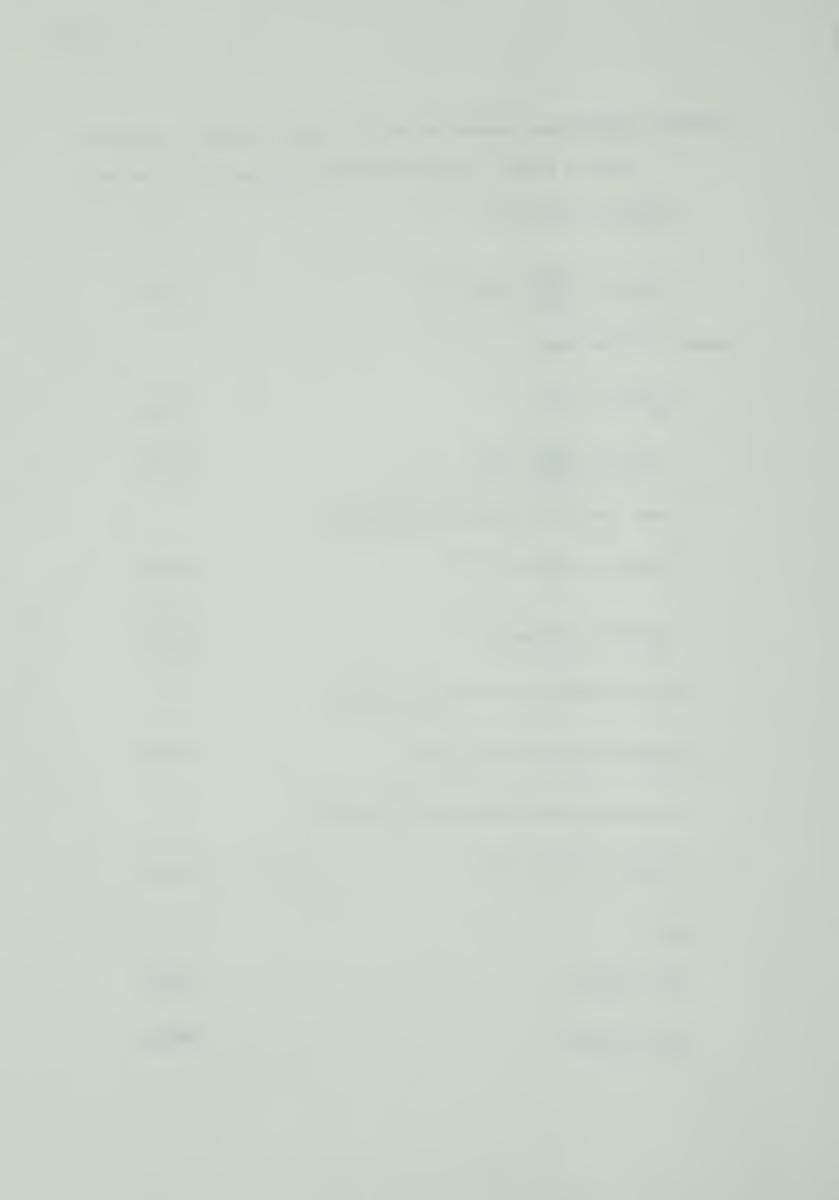
then using (B-16) and (B-17) one gets

$$W(x,s) = W(0,s) e^{-\gamma x}$$
 (B-19)

Let

$$W_{K} = W(0,s) \tag{B-20}$$

$$W_{K+1} = W(h,s)$$
 (B-21)



then

$$W_{K+1} = W_K \exp \left[-\gamma h\right] \tag{B-22}$$

Now we have by (B-7)

$$\gamma(s) = [s^2LC + (RC + GL) s + RG]^{\frac{1}{2}}$$
 (B-23)

then

$$\gamma(s) = s\sqrt{LC} \left[1 + \frac{1}{s} \left(\frac{R}{L} + \frac{G}{C}\right) + \frac{RG}{s^2 LC}\right]^{\frac{1}{2}}$$
 (B-24)

If the time t is small, then s is very large (s $\rightarrow \infty$ as t \rightarrow 0), then we can write

$$\gamma(s) = s\sqrt{LC} \left[1 + \frac{1}{s} \left(\frac{R}{L} + \frac{G}{C}\right)\right]^{\frac{1}{2}}$$
 (B-25)

Using the binomial theorem one gets

$$\gamma(s) = s\sqrt{LC} \left[1 + \frac{1}{2s} \left(\frac{R}{L} + \frac{G}{C}\right)\right]$$
 (B-26)

or

$$\gamma(s) = s\sqrt{LC} + \frac{1}{2} \left[\frac{R}{\rho} + G \rho \right]$$
 (B-27)

$$h.\gamma(s) = s.h.\sqrt{LC} + \frac{1}{2} \left[\frac{Rh}{\rho} + G h.\rho \right]$$
 (B-28)

Now by definition of Δt as the electrical length of the section with physical length h and letting

$$r = R.h (B-29)$$

$$g = G.h$$



we have

$$h.\gamma(s) = \Delta t.s + \frac{1}{2} \left[\frac{r}{\rho} + g \rho \right]$$
 (B-31)

Thus

$$W_{K+1} = W_K \exp \left[-\Delta t.s - \frac{1}{2} \left(\frac{r}{\rho} + g \rho \right) \right]$$
 (B-32)

which is equation (3-50)



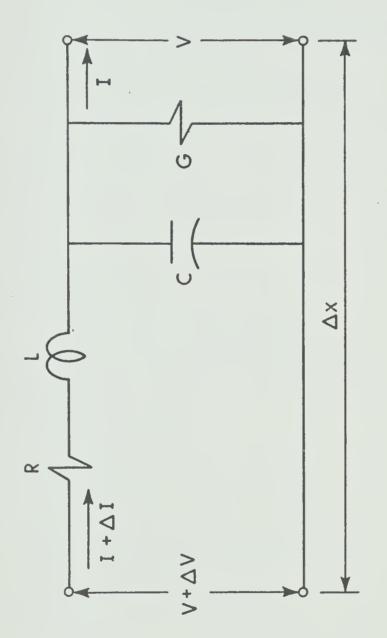


FIG. B-1 INCREMENTAL LENGTH OF A TRANSMISSION LINE



APPENDIX C

PROGRAM LISTING



```
*********** GENERAL TRANSMISSION LINE ***********
            C
                  ((K).V(K)...OLD CURRENTS AND VOLTAGES AT THE SECTION BOUNDARIES
            C
            C
                  CC(K), VV(K)... NEW VALUES CALCULATED FROM C(K), V(K)
                  LINE DIVIDED INTO NN-1 SECTIONS
ISN 0002
                  CIMENSICN YAB(1004), X11(1004), YBB(1004), X22(1004), YCB(1004), X33(10
                  2041
1SN 0003
                  CIMENSION C(1000).CC(1000).V(500).VV(500).X(1002.4),WORK(1024)
1SN 0004
                   DIMENSION FY(1001), Z(1001), DEL(1001), SY(1001), SZ(1002, 3)
ISN 0005
                  DIMENSION RU(1001) . YN(1002) . Z1(1002)
15N 0006
                   DIMENSION 7A(1002), COF1(501), COF2(501), COF5(501), COF6(501)
ISN 0007
                   DIMENSICA FC0(1002,4),Y(1002,4),Y3(1002,4),Y4(1002)
ISN 0008
                  DIMENSION RAB(104), RBB(504), RCB(1004), Z11(104), Z22(504), Z33(1004)
ISN 0009
                   EXTERNAL FL.FC.FR.FG.SOUR
15N 001C
                  COMMON/BGB/XLO.CO.RO.GO.XLE
ISN 0011
                   COMMON/SOS/XLE1.XLE2.XLE3.AMP.SLOPE
            C
                   WORK(N) ... . AUXILIARY VECTOR FOR PLOTTING
ISN 0012
                   CALL PLOTS (WORK(1),4096)
1SN 0013
                   (ALL PLOT (5.0.5.0.-3)
                   CUTPUT RESISTIVE LOAD=FR
1SN 0014
                   READ(5,901) RR.GG
ISN 0015
              901 FORMAT (2F16.8)
ISN 0016
                   READ(5.902) XNN. XMEW
ISN 0017
               902 FORMAT(2E16.8)
                   READ(5.903) RESI.ROVAL.GVAL
1SN 0018
               903 FORMAT(3E16.8)
ISN 0019
ISN 0020
                   READ(5.904) XL0.C0.R0.G0
1 SN 0 021
               904 FORMAT (4F16.8)
                   WRITE(6.910) XLO.CO.RO.GO
ISN 0022
ISN 0023
               910 FORMAT (4E10.4)
15N 0024
                  READ(5,905) XLE1.XLE2.XLE3.AMP.SLOPE
               905 FORMAT (5E12.4)
ISN 0025
                   IF (XMEW.EQ.0.0) GO TO 666
ISN 0026
                   ALFA=((RESI/ROVAL)+(GVAL*ROVAL))/2.
ISN 0028
ISN 0029
                   XNN=ALFA+SORT((ALFA)/(12.*XMFW))
                   WRITE(6.906) RESI, ROVAL, GVAL, XMEW, ALFA, XNN
ISN 0030
               906 FORMAT(1X.6E16.8)
ISN 0031
ISN 0032
                   NN=XNN
                   WRITE(6.907) NN
ISN 0033
ISN 0034
               907 FORMAT(30X,118)
                   IF (NN.LT.10) NN=10
ISN 0035
                   IF (NN.GT.1000) GO TO 777
1SN 0037
               666 FY(1)=0.00
ISN 0039
                   CO 1 K1=2 - 1001
ISN 0040
1SN 0041
                   DEL(1)=SGRT((FL(0.0))*(FC(0.0)))
ISN 0042
                   XK = K1
                   2(K1)=(XK-1.)*0.001
1SN 0043
                   2K = Z(K1)
15N 0044
                  DFL(K1)=SORT((FL(ZK))+(FC(ZK)))
ISN 0045
                1 FY(K1)=FY(K1-1)+0.5*(DEL(K1)+DEL(K1-1))
ISN 0046
```

TOU=FY(1001)-FY(1)

1SN 0C47



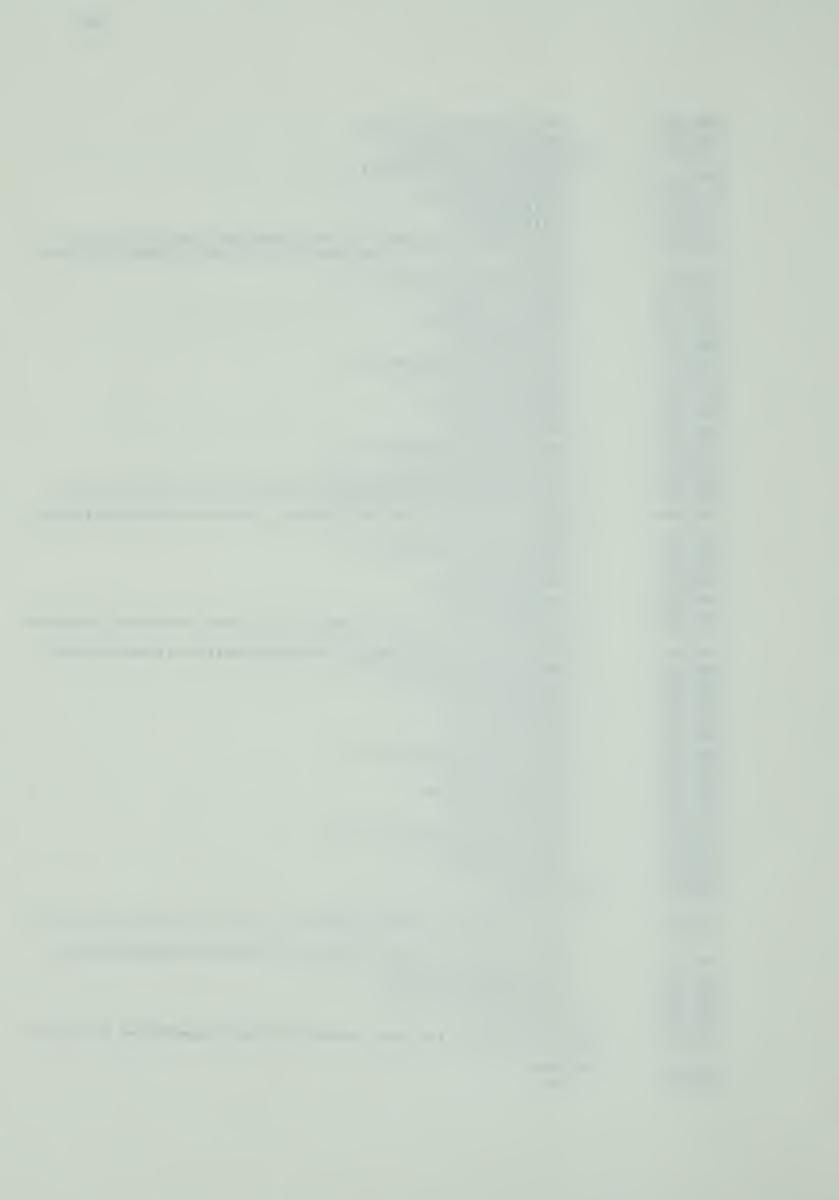
```
ISN 0048
                    DO 2 IX=1.1000
ISN 0049
                    I \times 1 = I \times + 1
ISN 0050
                  2 YN(IX)=FY(IX1)/TOU
ISN 0051
                    DO 3 IZ=1.1000
ISN 0052
                    IZ1= IZ+1
                  3 21(17)=2(171)
ISN 0053
ISN 0054
                    DO 7 INDEX=1.3
ISN 0055
                    IF ( INDEX .EQ. 2 ) XNN=1.5*XNN
ISN 0057
                    IF ( INDEX .EQ. 3 ) XNN=(4.0*XNN)/3.0
ISN 0059
                 10 DELT=TOU/XNN
ISN 0060
                    NN=XNN
ISN 0061
                    NP=NN-1
ISN 0062
                    DO 4 J1=1.NP
IŚN 0063
                    SZ(1.INDEX)=0.0
ISN 0064
                    XJ=J1
ISN 0065
                    SY(J1)=XJ*DELT
ISN 0066
                    DO 4 K2=1.1000
ISN 0067
                    IF(SY(J1) .GT. FY(K2+1)) GC TO 4
ISN 0069
                    IF(SY(J1) *LT* FY(K2+1) *AND* SY(J1)*GT* FY(K2)) Y1=FY(K2)
1SN 0071
                    IF(SY(J1) .LT. FY(K2+1) .AND. SY(J1).GT. FY(K2)) Y6=FY(K2+1)
ISN 0073
                    IF(SY(J1) .LT. FY(K2+1) .AND. SY(J1).GT. FY(K2)) KNN=K2
ISN 0075
                    J111=J1+1
1SN 0076
                    SZ(J111 .INDEX) = Z(KNN) + (0.001 * ((SY(J1) - Y1) / (Y6 - Y1)))
ISN 0077
                  4 CONTINUE
1SN 0078
                    SZ(NN+1.INDEX)=1.0
1SN 0079
                    NX=NN+1
ISN 0080
                    NY=NN+2
ISN 0081
                    CO 5 J5=1.NX
ISN 0082
                    XZJ=SZ(J5.INDEX)
ISN 0083
                    XL=FL(XZJ)
ISN 0084
                    XC = FC(XZJ)
                    XR=FR(XZJ)
ISN 0085
ISN 0086
                    XG=FG(XZJ)
                    PO(J5)=SQRT(XL/XC)
ISN 0087
                    COF1(J5)=1.-(DELT*XG)/(2.*XC)
ISN 0088
ISN 0089
                    CDF2(J5)=RP(J5)-(DELT*XR)/(2.*(SQRT(XL*XC)))
ISN 0090
                    CDF5(J5) = (2.)*(1.+(DELT*XG)/(2.*XC))
                    CDF6(J5)=(2*)*(RO(J5)+(DELT*XR)/(2**(SQRT(XL*XC))))
1SN 0091
ISN 0092
                  5 F00(J5.INDEX)=R0(J5)
             C *** INITIAL CONDITIONS ON TRANSMISSION LINE ***
1SN 0093
                    DO 6 K3=1.NN
15N 0094
                    C(K3)=0.0
ISN 0095
                  6 V(K3)=0.0
             C *** ITERATIVE CALCULATION, L=DISCRETE TIME ***
1SN 0096
                    NH=5*NN
ISN 0057
                    DC 7 L=1.NH
                    XLE=L/XNN
15N 0058
ISN 0099
                    NP=NN-1
                    DO 8 J=2.NP
ISN 0100
                    A=V(J-1)*COF1(J-1)+C(J-1)*COF2(J-1)
ISN 0101
                    P=V(J+1)*CNF1(J+1)-C(J+1)*CDF2(J+1)
ISN 0102
                    VV(J) = (A+R) /CCF5(J)
ISN 0103
                    CC(J)=(A-P)/COF6(J)
15N 0104
                 & CONTINUE
ISN 0105
             C *** POUNDARY CONDITIONS ***
                    XA=0.5*(COF5(1)+GG*COF6(1))
ISN 0106
```



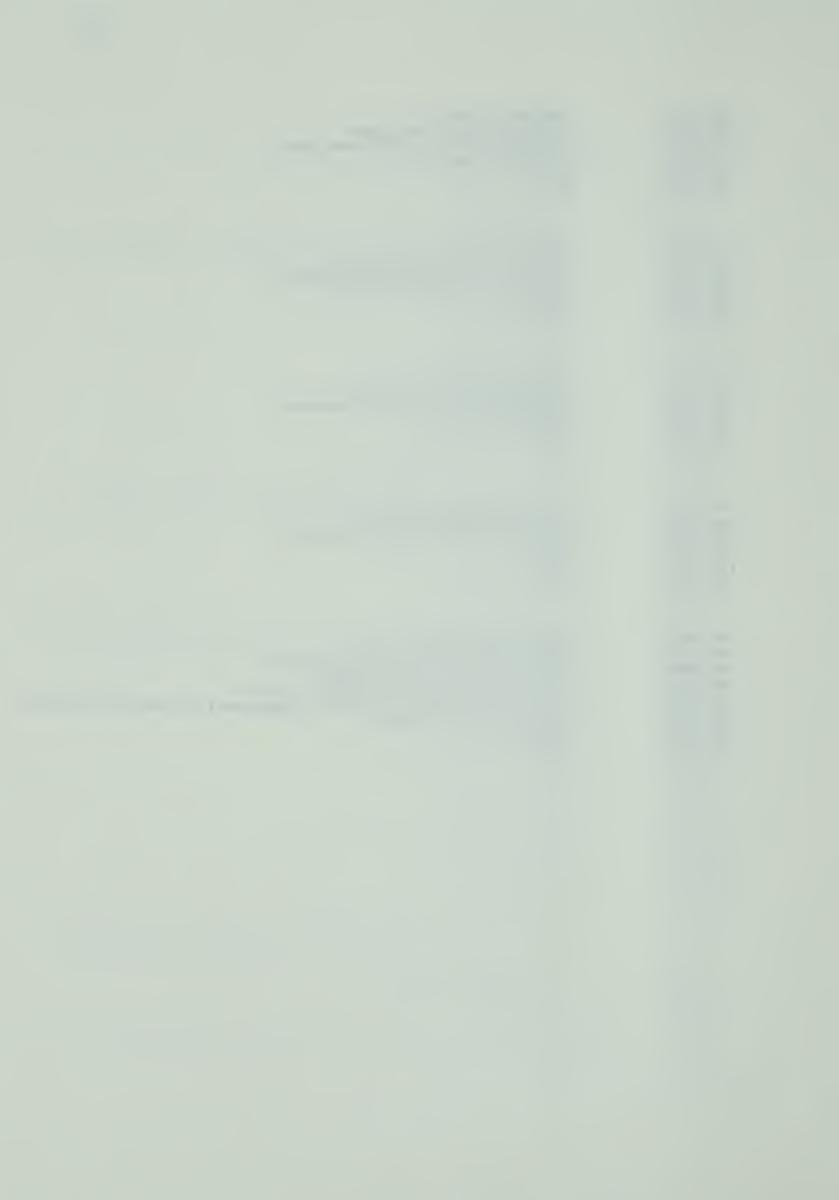
```
ISN 0107
                    VV(1) = (0.5*(SDUR(XLE)*CDF6(1))+(V(2)*CDF1(2))-(C(2)*CDF2(2)))/XA
ISN 0108
                    CC(1)=SOUR(XLE)-GG*VV(1)
                    XB=0.5*(CDF5(NN)+(CDF6(NN)/RR))
1SN 0109
ISN 0110
                    VV(NN)=((V(NP)*COF1(NP))+(C(NP)*COF2(NP)))/XB
15N 0111
                    CC(NN)=VV(NN)/RR
             C *** OLD - NEW VALUES EXCHANGE ***
ISN 0112
                    00 9 M=1.NN
ISN 0113
                    CIM) = CC(M)
ISN 0114
                 9 V(M)=VV(M)
ISN 0115
                444 X(L.INDEX)=XLE
ISN 0116
                    Y(L.INDEX)=C(NN)
ISN 0117
                    Y3(L, INDEX) = SOUR (XLE)
                    PRINT THE CURRENTS AT THE EVERY FIFTH TIME INSTANT:
15N 0118
                    IF(L/5*5.EO.L) WRITE (6.908) C(1),C(NN/4).C(NN/2),C(3*NN/4),C(NN)
                908 FCRMAT (1X.6E20.7)
ISN 0120
15N 0121
                  7 CONTINUE
ISN 0122
                    RB11=0.0
1SN 0123
                    YB11=0.0
1SN 0124
                    NAX=(5*NN)/2
                    00 71 L=1.NAX
ISN 0125
                    YAB(L)=Y(L.1)
ISN 0126
                    Y811=AMAX1 (YA8(L), Y811)
1SN 0127
ISN 0128
                 71 X11(L)=X(L.1)
                    1CN=NN/2
ISN 0129
ISN 0130
                    DD 76 L=1.1CN
                    RAB(L)=ROC(L.1)
15N 0131
                    RR11=AMAX1 (RAB(L), RB11)
1SN 0132
                    211(L)=S2(L.1)
1SN 0133
                 76 CENTINUE
1SN 0134
                    YC11=0.0
ISN 0135
                    RC11=0.0
1SN 0136
                    NBX=(15*NN)/4
ISN 0137
                    DD 72 L=1 .NBX
1SN 0138
                    YBB(L)=Y(L.2)
ISN 0139
                    YC11=AMAX1(YPE(L),YC11)
1SN 0140
                 72 ×22(L)=X(L,2)
ISN 0141
ISN 0142
                    JCN={3*NN}/4
                    CO 77 L=1.JCN
1SN 0143
                    FR8(L)=R00(L.2)
ISN 0144
                    RC11=AMAX1 (R88(L) . RC11)
15N 0145
                    222(L)=S2(L.2)
ISN 0146
                 77 CUNTINUE
1SN 0147
                    YD11=0.0
ISN 0148
                    RD11=0.0
1SN 0145
                    NCX=5*NN
ISN 0150
ISN 0151
                    00 73 L=1.NCX
15N 0152
                    YCE(L)=Y(L.3)
                    YD11=AMAX1(YCB(L),YDI1)
ISN 0153
                 73 X33(L)=X(L.3)
1SN 0154
                    KON=NN
15N 0155
                    TC: 78 L=1.KCN
ISN 0156
                    RCB(L)=RGO(L.3)
1SN 0157
                    FD[]=AMAX1(RCB(L),ROI1)
1SN 0158
                    233(L)=SZ(L+3)
ISN 0159
                 78 CONTINUE
15N 016C
                    YRIG=AMAXI(YRII.YCII.YDII)
ISN 0161
```



```
ISN 0162
                    RBIG=AMAX1(RBII,RC11,RDI1)
ISN 0163
                    WRITF(6.74) YBIG.RBIG
                74 FORMAT(30X.2E16.8./)
ISN 0164
             C *** PLOTTING INSTRUCTIONS ***
ISN 0165
                    RAB(ION+1)=0.0
ISN 0166
                    FAB(ION+2)=RBIG/4.
ISN 0167
                    711(ION+1)=0.0
ISN 0168
                    711(ION+2)=0.2
15N 0169
                    CALL AXIS(0.0.0.0. RC. .2.4.0.90.0. RAB([ON+1].RAR([ON+2].20.0)
ISN 0170
                    CALL AX1S(0.0.0.0.*ACTUAL LENTH*.-13.5.0.0.0.ZI1(ION+I).ZI1(1ON+2)
                   1.20.0)
ISN 0171
                    CALL LINE(ZII.RAB.1CN.I.0.3)
                    CALL PLOT (0.0.0.0.-3)
ISN 0172
ISN 0173
                    0.0=(I+NOL)889
1SN 0174
                    PPB(JON+2)=RB1G/4.
ISN 0175
                    722(JON+1)=0.0
ISN 0176
                    722(JON+2)=0.2
ISN 0177
                    CALL LINE(222.RBB.JON.1.0.3)
ISN 0178
                    CALL PLOT (0.0.0.0.-3)
ISN 0179
                    RCB(KON+1)=0.0
ISN 0180
                    RCB(KON+2)=RBIG/4.
ISN OTEL
                    233(KON+1)=0.0
1SN 0182
                    233(KON+2)=0,2
ISN 0183
                    CALL LINE(233,RCB,KON,1,0,3)
1SN 0184
                    (ALL PLOT (0.0,5.5,-3)
ISN 0185
                    CALL SCALE (YN.4.0.1000.1.20.0)
1SN 0186
                    CALL AXIS (0.0,0.0, "YN(1X)",6,4.0,90.0, YN(1001), YN(1002),20.0)
ISN 0187
                    CALL SCALE(21.5.0.1000.1.20.0)
ISN 0188
                    CALL AXIS (0.0.0.0.*ACTUAL LENGTH*,-13.5.0.0.0.21(1001).21(1002).2
                   10.01
ISN 0185
                    CALL LINE (21.YN.1000.1.0.3)
                    CALL PLOT (12.0.0.0.-3)
ISN 0190
                    YAE(NAX+1)=0.0
ISN 0191
ISN 0192
                    YAB(NAX+2)=YBIG/4.
ISN 0193
                    YII(NAX+I)=0.0
                    X11(NAX+2)=1.0
ISN 0194
                    CALL AXIS (0.0.0.0.*LOAD CUPRENT*.12.4.0,90.0.YAB(NAX+1).YAB(NAX+2
ISN 0195
                   1).20.0)
                    CALL AXIS(0.0.0.0.*TIME*.-4.5.0.0.0.XII(NAX+I).X11(NAX+2).20.0)
ISN 0196
ISN 0197
                    CALL LINE (XII.YAB.NAX.I.0.3)
ISN 0198
                    CALL PLOT (0.0.0.0.-3)
ISN 0199
                    YBB(NBX+1)=0.0
ISN 0200
                    YPB(NDX+2)=YBIG/4.
ISN 0201
                    x22(NRX+1)=0.0
ISN 0202
                    x22(NBX+2)=1.0
                    CALL LINE (X22.YBB.NBX.1.0.3)
ISN 0203
                    CALL PLOT (0.0.0.0.-3)
ISN 0204
                    YC9(NCX+1)=0.0
ISN 0205
ISN 0206
                    YCB(NCX+2)=YBIG/4
ISN 0207
                    x33(NCX+1)=0.0
                    ¥33(NCX+2)=1.0
ISN 0208
                    CALL LINE (X33, YCB, NCX, 1.0, 3)
ISN 0209
                    CALL PLOT (0.0.5.5.-3)
ISN 0210
                    CO 600 I=I.1000
ISN 0211
                    Y4(1)=Y3(1.3)
1 SN 0212
               600 CONTINUE
ISN 0213
                    CALL SCALF (Y4.4.0.1000.1.20.0)
ISN 0214
                    CALL AXIS(0.0.0.0.*!NPUT CURRENT*,13.4.0.90.7, YA(1001), Y4(1002).20
ISN 0215
                   1.0)
                    CALL AXIS(0.0.0.0.1 TIME 1.-4.5.0.0.0.1 X33(NCX+1).X33(NCX+2).20.0)
ISN 0216
                    CALL LINE(X33.Y4.1000.1.0.3)
ISN 0217
                    CALL PLOT (10.0.7.0.-3)
ISN 021E
                    CALL PLOT (0.0.0.0.999)
ISN 0219
ISN 0220
                    CO TO 888
               777 WELTE (6.909) NN
ISN 0221
                509 FORMAT(30X,49H FOR THESE VALUES OF SYSTEM PARAMETERS WE GET NN=,18
ISN 0222
                  1/)
ISN 0223
                989 STOP
                   END
ISN 0224
```



I SN I SN I SN I SN	0002 0003 0004 0005 0006	FUNCTION FC(DAN) CCMMGN/BCP/XLO.CO.RO.GO.XLF CCMMGN/SOS/XLF1.XLE2.XLE3.AMP.SLCPE FC=CO*EXP(-CAN) FETURN END
ISN	2002	FUNCTION FL (DAN)
	0003	CCMMON/BOB/XLO,CO.RO.GO,XLE
1SN	0004	COMMON/SOS/XLE1.XLE2.XLE3.AMP.SLOPE
ISN	0005	FL=XL0*(2.0+SIN(22.0*DAN/7.0))
ISN	0006	FETURN
ISŅ	0007	END
	0002	FUNCTION FF (DAN)
	0003	CCMMCN/BCP/XL0,C0,R0,G0,XLE
	0004	CCMMON/SGS/XLEI,XLE2,XLE3,AMF,SLOPE
	0005	FR=R0*ALOG(1.0+DAN)
	0006	FTURN FAMOUR FOR THE PROPERTY OF THE PROPERTY
150	0007	END
ISN	0002	FUNCTION FG(DAN)
ISN	0003	CCMMCN/BCP/XLO,CO,RO,GO,XLE
ISN	0004	COMMON/SOS/XLFI, XLE2, XLE3, AMP, SECPE
	0005	F G= G0
ISN	0 € 0 €	FETUPN
ISN	0007	END
ISN	0002	FUNCTION SOUR(XAX)
N2 1	0003	COMMON/BOB/XL0.C0.R0.G0.XLE
ISN	0 0 0 4	COMMON/SOS/XLEI, XLEZ, XLEZ, AMP, SLUPE
ISN	0005	IF(XAX.LT.XLEI) SDUR=SLOPE*XAX
	0007	IF(XAX.LT.XLE2.AND.XAX.GT.XLE1) SOUR=AMP
ISN	0009	IF(XAX.LT.YLE3.AND.XAX.GT.XLF2) SOUR=AMP*((XLE3-XAX)/(XLE3-XLE2))
	0011	IF(XAX.GT.XLE3) SOUR=0.0
	0013	HETURN
ISN	0014	FND









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